# 1Gb Q-die DDR2 1.55V SDRAM Specification

60FBGA with Lead-Free & Halogen-Free (RoHS compliant)

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# **Revision History**

Revision	Month	Year	History
1.0	April	2008	- Initial Release
1.01	May	2008	- typo correction
1.1	July	2008	- Updated AC timing table with the JEDEC update(JESD79-2E)
1.2	December	2008	- Updated AC/DC operating condition with the JEDEC update(JESD79-2E)

# 1.0 Ordering Information

Org.	DDR2-667 5-5-5	Package
256Mx4	K4T1G044QQ-HYE6	60 FBGA
128Mx8	K4T1G084QQ-HYE6	60 FBGA

#### Note:

- 1. Speed bin is in order of CL-tRCD-tRP.
- 2. "H" of Part number(12th digit) stands for Lead-Free, Halogen-Free, and RoHS compliant products.
- 3. "Y" of Part number(13th digit) stands for Low Voltage product.

### 2.0 Key Features

Speed	DDR2-667 5-5-5	Units
CAS Latency	5	tCK
tRCD(min)	15	ns
tRP(min)	15	ns
tRC(min)	60	ns

- JEDEC Compliant V<sub>DD</sub> = 1.55V ± 0.05V Power Supply
- $V_{DDQ} = 1.55V \pm 0.05V$
- Backward compatible with DDR2 1.8V ± 0.1V
- 333MHz f<sub>CK</sub> for 667Mb/sec/pin
- · 8 Banks
- Posted CAS
- Programmable CAS Latency: 3, 4, 5
- · Programmable Additive Latency: 0, 1, 2, 3, 4
- Write Latency(WL) = Read Latency(RL) -1
- Burst Length: 4, 8(Interleave/Nibble sequential)
- Programmable Sequential / Interleave Burst Mode
- Bi-directional Differential Data-Strobe (Single-ended datastrobe is an optional feature)
- · Off-Chip Driver(OCD) Impedance Adjustment
- On Die Termination
- · Special Function Support
  - 50ohm ODT
  - High Temperature Self-Refresh rate enable
- Average Refresh Period 7.8us at lower than T<sub>CASE</sub> 85°C, 3.9us at 85°C < T<sub>CASE</sub> ≤ 95 °C
- All of products are Lead-Free, Halogen-Free, and RoHS compliant

The 1Gb DDR2 SDRAM is organized as a 32Mbit x 4 I/Os x 8banks, 16Mbit x 8 I/Os x 8banks device. This synchronous device achieves high speed double-data-rate transfer rates of 667Mb/sec/pin (DDR2-667) for general applications.

The chip is designed to comply with the following key DDR2 SDRAM features such as posted CAS with additive latency, write latency = read latency - 1, Off-Chip Driver(OCD) impedance adjustment and On Die Termination.

All of the control and address inputs are synchronized with a pair of externally supplied differential clocks. Inputs are latched at the crosspoint of differential clocks (CK rising and  $\overline{CK}$  falling). All I/Os are synchronized with a pair of bidirectional strobes (DQS and  $\overline{DQS}$ ) in a source synchronous fashion. The address bus is used to convey row, column, and bank address information in a  $\overline{RAS}$ /  $\overline{CAS}$  multiplexing style. For example, 1Gb(x8) device receive 14/ 10/3 addressing.

The 1Gb DDR2 device operates with a single  $1.55V \pm 0.05V$  power supply and  $1.55V \pm 0.05V$  V<sub>DDQ</sub>.

The 1Gb DDR2 device is available in 60ball FBGAs(x4/x8).

Note: The functionality described and the timing specifications included in his data sheet are for the DLL Enabled mode of operation.

Note: This data sheet is an abstract of full DDR2 specification and does not cover the common features which are described in "DDR2 SDRAM Device Operation & Timing Diagram".

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# 3.0 Package Pinout/Mechanical Dimension & Addressing

### 3.1 x4 package pinout (Top View) : 60ball FBGA Package

Α	V <sub>DD</sub>	NC	V <sub>SS</sub>	$V_{SSQ}$	DQS	$V_{\mathrm{DDQ}}$
В	NC	$V_{SSQ}$	DM	DQS	$V_{SSQ}$	NC
С	V <sub>DDQ</sub>	DQ1	$V_{DDQ}$	$V_{\mathrm{DDQ}}$	DQ0	$V_{DDQ}$
D	NC	V <sub>SSQ</sub>	DQ3	DQ2	V <sub>SSQ</sub>	NC

4 5 6

V<sub>DDL</sub> V<sub>SS</sub>  $\mathbf{V}_{\mathsf{REF}}$ F CKE WE G BA2 BA0 BA1 Н A10/AP Α1  $V_{SS}$ А3 Α5 Κ Α7 Α9  $V_{DD}$ A12 NC

V <sub>SSQ</sub>	DQS	$V_{DDQ}$
DQS	$V_{SSQ}$	NC
$V_{\mathrm{DDQ}}$	DQ0	$V_{DDQ}$
DQ2	$V_{SSQ}$	NC
V <sub>SSDL</sub>	СК	$V_{DD}$
RAS	СК	ODT0
CAS	cs	
A2	A0	$V_{DD}$
A6	A4	
A11	A8	V <sub>SS</sub>
NC	A13	

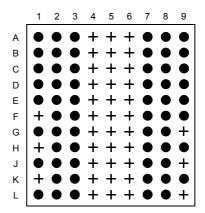
Note :  $V_{DDL}$  and  $V_{SSDL}$  are power and ground for the DLL. It is recommended that they be isolated on the device from  $V_{DD}$ ,  $V_{DDQ}$ ,  $V_{SS}$ , and  $V_{SSQ}$ .

# **Ball Locations (x4)**

- Populated ball
- + Ball not populated

Top view

(See the balls through the package)



# 3.2 x8 package pinout (Top View): 60ball FBGA Package

1 2 3 4 5 6 7 8 9
-------------------

Α	V <sub>DD</sub>	NU/RDQS	V <sub>SS</sub>
В	DQ6	V <sub>SSQ</sub>	DM/RDQS
С	V <sub>DDQ</sub>	DQ1	$V_{\mathrm{DDQ}}$
D	DQ4	$V_{SSQ}$	DQ3
Е	V <sub>DDL</sub>	V <sub>REF</sub>	V <sub>SS</sub>
F		CKE	WE
G	BA2	BA0	BA1
н		A10/AP	A1
J	V <sub>SS</sub>	А3	<b>A</b> 5
K		A7	A9
L	V <sub>DD</sub>	A12	NC

V <sub>SSQ</sub>	DQS	$V_{DDQ}$
DQS	$V_{SSQ}$	DQ7
$V_{\mathrm{DDQ}}$	DQ0	$V_{DDQ}$
DQ2	$V_{SSQ}$	DQ5
V <sub>SSDL</sub>	СК	$V_{DD}$
RAS	CK	ODT0
CAS	cs	
A2	A0	$V_{DD}$
A6	A4	
A11	A8	V <sub>SS</sub>
NC	A13	

#### Note:

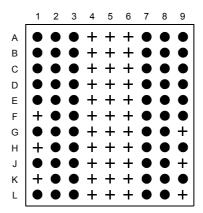
- 1. Pins B3 and A2 have identical capacitances as pins B7 and A8.
- 2. For a Read, when enabled, strrobe pair RDQS & RDQS are identical in function and timing to strobe pair DQS & DQS and input data masking function is disabled.
- 3. The function of DM or RDQS/RDQS is enabled by EMRS command.
- 4.  $V_{DDL}$  and  $V_{SSDL}$  are power and ground for the DLL. It is recommended that they be isolated on the device from  $V_{DD}$ ,  $V_{DDQ}$ ,  $V_{SS}$ , and  $V_{SSQ}$ .

# **Ball Locations (x8)**

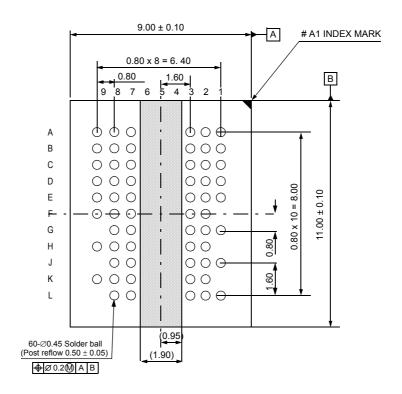
- Populated ball
- + Ball not populated

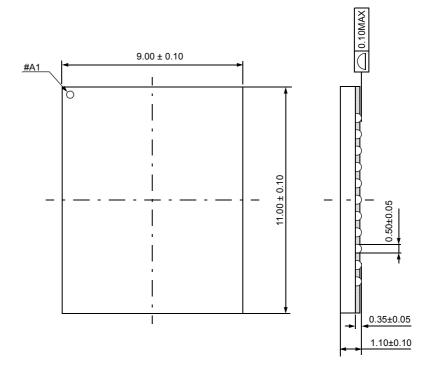
#### Top view

(See the balls through the package)



### 3.3 FBGA Package Dimension (x4/x8)





# 4.0 Input/Output Functional Description

Symbol	Туре	Function
CK, CK	Input	Clock: CK and $\overline{\text{CK}}$ are differential clock inputs. All address and control input signals are sampled on the crossing of the positive edge of CK and negative edge of $\overline{\text{CK}}$ . Output (read) data is referenced to the crossings of CK and $\overline{\text{CK}}$ (both directions of crossing).
CKE	Input	Clock Enable: CKE HIGH activates, and CKE Low deactivates, internal clock signals and device input buffers and output drivers. Taking CKE Low provides Precharge Power-Down and Self Refresh operation (all banks idle), or Active Power-Down (row Active in any bank). CKE is synchronous for power down entry and exit, and for self refresh entry. CKE is asynchronous for self refresh exit. After V <sub>REF</sub> has become stable during the power on and initialization swquence, it must be maintained for proper operation of the CKE receiver. For proper self-refresh entry and exit, V <sub>REF</sub> must be maintained to this input. CKE must be maintained high throughout read and write accesses. Input buffers, excluding CK, CK, ODT and CKE are disabled during power-down. Input buffers, excluding CKE, are disabled during
CS	Input	self refresh.  Chip Select: All commands are masked when $\overline{CS}$ is registered HIGH. $\overline{CS}$ provides for external Rank selection on systems with multiple Ranks. $\overline{CS}$ is considered part of the command code.
ODT	Input	On Die Termination: ODT (registered HIGH) enables termination resistance internal to the DDR2 SDRAM. When enabled, ODT is only applied to each DQ, DQS, DQS, RDQS, RDQS, and DM signal for x4/x8 configurations. For x16 configuration ODT is applied to each DQ, UDQS/UDQS, LDQS/LDQS, UDM, and LDM signal. The ODT pin will be ignored if the Extended Mode Register (EMRS(1)) is programmed to disable ODT.
RAS, CAS, WE	Input	Command Inputs: RAS, CAS and WE (along with CS) define the command being entered.
DM	Input	Input Data Mask: DM is an input mask signal for write data. Input data is masked when DM is sampled HIGH coincident with that input data during a Write access. DM is sampled on both edges of DQS. Although DM pins are input only, the DM loading matches the DQ and DQS loading. For x8 device, the function of DM or RDQS/RDQS is enabled by EMRS command.
BA0 - BA2	Input	<b>Bank Address Inputs:</b> BA0, BA1 and BA2 define to which bank an Active, Read, Write or Precharge command is being applied. Bank address also determines if the mode register or extended mode register is to be accessed during a MRS or EMRS cycle.
A0 - A13	Input	Address Inputs: Provided the row address for Active commands and the column address and Auto Precharge bit for Read/Write commands to select one location out of the memory array in the respective bank. A10 is sampled during a Precharge command to determine whether the Precharge applies to one bank (A10 LOW) or all banks (A10 HIGH). If only one bank is to be precharged, the bank is selected by BA0, BA1. The address inputs also provide the op-code during Mode Register Set commands.
DQ	Input/Out- put	Data Input/ Output: Bi-directional data bus.
DQS, ( <u>DQS)</u> (LDQS), ( <u>LDQS)</u> (UDQS), ( <u>UDQS</u> ) (RDQS), (RDQS)	Input/Out- put	Data Strobe: output with read data, input with write data. Edge-aligned with read data, centered in write data. For the x16, LDQS corresponds to the data on DQ0-DQ7; UDQS corresponds to the data on DQ8-DQ15. For the x8, an RDQS option using DM pin can be enabled via the EMRS(1) to simplify read timing. The data strobes DQS, LDQS, UDQS, and RDQS may be used in single ended mode or paired with optional complementary signals DQS, LDQS, UDQS, and RDQS to provide differential pair signaling to the system during both reads and writes. An EMRS(1) control bit enables or disables all complementary data strobe signals.  In this data sheet, "differential DQS signals" refers to any of the following with A10 = 0 of EMRS(1)
NC		No Connect: No internal electrical connection is present.
$V_{DD}/V_{DDQ}$	Supply	Power Supply: 1.55V +/- 0.05V, DQ Power Supply: 1.55V +/- 0.05V
$V_{SS}/V_{SSQ}$	Supply	Ground, DQ Ground
$V_{DDL}$	Supply	DLL Power Supply: 1.55V +/- 0.05V
$V_{SSDL}$	Supply	DLL Ground
$V_{REF}$	Supply	Reference voltage

# 5.0 DDR2 SDRAM Addressing

# 1Gb Addressing

Configuration	256Mb x4	128Mb x 8	64Mb x16
# of Bank	8	8	8
Bank Address	BA0 ~ BA2	BA0 ~ BA2	BA0 ~ BA2
Auto precharge	A10/AP	A10/AP	A10/AP
Row Address	A0 ~ A13	A0 ~ A13	A0 ~ A12
Column Address	A0 ~ A9,A11	A0 ~ A9	A0 ~ A9

<sup>\*</sup> Reference information: The following tables are address mapping information for other densities.

#### 256Mb

Configuration	64Mb x4	32Mb x 8	16Mb x16
# of Bank	4	4	4
Bank Address	BA0,BA1	BA0,BA1	BA0,BA1
Auto precharge	A10/AP	A10/AP	A10/AP
Row Address	A0 ~ A12	A0 ~ A12	A0 ~ A12
Column Address	A0 ~ A9,A11	A0 ~ A9	A0 ~ A8

#### 512Mb

Configuration	128Mb x4	64Mb x 8	32Mb x16
# of Bank	4	4	4
Bank Address	BA0,BA1	BA0,BA1	BA0,BA1
Auto precharge	A10/AP	A10/AP	A10/AP
Row Address	A0 ~ A13	A0 ~ A13	A0 ~ A12
Column Address	A0 ~ A9,A11	A0 ~ A9	A0 ~ A9

#### 2Gb

Configuration	512Mb x4	256Mb x 8	128Mb x16
# of Bank	8	8	8
Bank Address	BA0 ~ BA2	BA0 ~ BA2	BA0 ~ BA2
Auto precharge	A10/AP	A10/AP	A10/AP
Row Address	A0 ~ A14	A0 ~ A14	A0 ~ A13
Column Address	A0 ~ A9,A11	A0 ~ A9	A0 ~ A9

# 4Gb

Configuration	1 Gb x4	512Mb x 8	256Mb x16
# of Bank	8	8	8
Bank Address	BA0 ~ BA2	BA0 ~ BA2	BA0 ~ BA2
Auto precharge	A10/AP	A10/AP	A10/AP
Row Address	A0 - A15	A0 - A15	A0 - A14
Column Address/page size	A0 - A9,A11	A0 - A9	A0 - A9



# 6.0 Absolute Maximum DC Ratings

Symbol	Parameter	Rating	Units	Notes
$V_{\mathrm{DD}}$	Voltage on V <sub>DD</sub> pin relative to V <sub>SS</sub>	- 1.0 V ~ 2.3 V	V	1
$V_{DDQ}$	Voltage on V <sub>DDQ</sub> pin relative to V <sub>SS</sub>	- 0.5 V ~ 2.3 V	V	1
$V_{DDL}$	Voltage on V <sub>DDL</sub> pin relative to V <sub>SS</sub>	- 0.5 V ~ 2.3 V	V	1
$V_{IN,}V_{OUT}$	Voltage on any pin relative to V <sub>SS</sub>	- 0.5 V ~ 2.3 V	V	1
T <sub>STG</sub>	Storage Temperature	-55 to +100	°C	1, 2

#### Note:

- 1. Stresses greater than those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.
- 2. Storage Temperature is the case surface temperature on the center/top side of the DRAM. For the measurement conditions, please refer to JESD51-2 standard.
- 3.  $V_{DD}$  and  $V_{DDQ}$  must be within 300mV of each other at all times; and  $V_{REF}$  must be not greater than 0.6 x  $V_{DDQ}$ . When  $V_{DD}$  and  $V_{DDQ}$  and  $V_{DDQ}$  are less than 500mV,  $V_{REF}$  may be equal to or less than 300mV.
- 4. Voltage on any input or I/O may not exceed voltage on V<sub>DDQ</sub>.

# 7.0 AC & DC Operating Conditions

#### 7.1 Recommended DC Operating Conditions (SSTL - 1.8)

Cumbal	Dovernator	Rating			Unito	Notes
Symbol	Parameter	Min.	Тур.	Max.	Units	Notes
$V_{DD}$	Supply Voltage	1.5	1.55	1.6	V	
V <sub>DDL</sub>	Supply Voltage for DLL	1.5	1.55	1.6	V	4
$V_{\mathrm{DDQ}}$	Supply Voltage for Output	1.5	1.55	1.6	V	4
V <sub>REF</sub>	Input Reference Voltage	0.49*V <sub>DDQ</sub>	0.50*V <sub>DDQ</sub>	0.51*V <sub>DDQ</sub>	mV	1,2
V <sub>TT</sub>	Termination Voltage	V <sub>REF</sub> -0.04	V <sub>REF</sub>	V <sub>REF</sub> +0.04	V	3

Note: There is no specific device  $V_{DD}$  supply voltage requirement for SSTL-1.8 compliance. However under all conditions  $V_{DDQ}$  must be less than or equal to  $V_{DD}$ .

- 1. The value of V<sub>REF</sub> may be selected by the user to provide optimum noise margin in the system. Typically the value of V<sub>REF</sub> is expected to be about 0.5 x V<sub>DDQ</sub> of the transmitting device and V<sub>REF</sub> is expected to track variations in V<sub>DDQ</sub>.
- 2. Peak to peak AC noise on V<sub>REF</sub> may not exceed +/-2% V<sub>REF</sub>(DC).
- 3. V<sub>TT</sub> of transmitting device must track V<sub>REF</sub> of receiving device.
- 4. AC parameters are measured with  $V_{DD},\,V_{DDQ}$  and  $V_{DDL}$  tied together.
- 5. It is backward compatible with DDR2 1.8V +/- 0.1V power supply.



#### 7.2 Operating Temperature Condition

Symbol	Symbol Parameter		Units	Notes	
T <sub>OPER</sub>	Operating Temperature	0 to 95	°C	1, 2	

- 1. Operating Temperature is the case surface temperature on the center/top side of the DRAM. For the measurement conditions, please refer to JESD51.2 standard.
- 2. At 85 95 °C operation temperature range, doubling refresh commands in frequency to a 32ms period (tREFI=3.9 us) is required, and to enter to self refresh mode at this temperature range, an EMRS command is required to change internal refresh rate.

#### 7.3 Input DC Logic Level

Symbol	Parameter	Min.	Max.	Units	Notes
V <sub>IH</sub> (DC)	DC input logic high	V <sub>REF</sub> + 0.125	V <sub>DDQ</sub> + 0.3	V	
V <sub>IL</sub> (DC)	DC input logic low	- 0.3	V <sub>REF</sub> - 0.125	V	

#### 7.4 Input AC Logic Level

Complete I	Downwood out	DDR	Haita	
Symbol	Parameter	Min.	Max.	Units
V <sub>IH</sub> (AC)	AC input logic high	V <sub>REF</sub> + 0.200	V <sub>DDQ</sub> + V <sub>PEAK</sub>	V
V <sub>IL</sub> (AC)	AC input logic low	V <sub>SSQ</sub> - V <sub>PEAK</sub>	V <sub>REF</sub> - 0.200	V

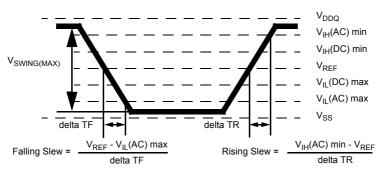
#### Note:

#### 7.5 AC Input Test Conditions

Symbol	Condition	Value	Units	Notes
$V_{REF}$	Input reference voltage	0.5 * V <sub>DDQ</sub>	V	1
V <sub>SWING(MAX)</sub>	Input signal maximum peak to peak swing	1.0	V	1
SLEW	Input signal minimum slew rate	1.0	V/ns	2, 3

#### Note:

- 1. Input waveform timing is referenced to the input signal crossing through the V<sub>IH/IL</sub>(AC) level applied to the device under test.
- 2. The input signal minimum slew rate is to be maintained over the range from V<sub>REF</sub> to V<sub>IH</sub>(AC) min for rising edges and the range from V<sub>REF</sub> to V<sub>IL</sub>(AC) max for falling edges as shown in the below figure.
- 3. AC timings are referenced with input waveforms switching from  $V_{IL}(AC)$  to  $V_{IH}(AC)$  on the positive transitions and  $V_{IH}(AC)$  to  $V_{IL}(AC)$  on the negative transitions.



< AC Input Test Signal Waveform >

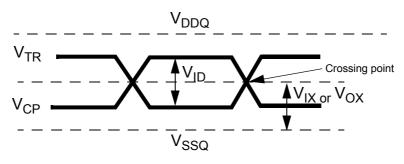
For information related to V<sub>PEAK</sub> value, Refer to overshoot/undershoot specification in device operation and timing datasheet; maximum peak amplitude allowed for overshoot and undershoot.

#### 7.6 Differential input AC logic Level

Symbol	Parameter	Min.	Max.	Units	Notes
V <sub>ID</sub> (AC)	AC differential input voltage	0.5	$V_{\mathrm{DDQ}}$	V	1
V <sub>IX</sub> (AC)	AC differential cross point voltage	0.5 * V <sub>DDQ</sub> - 0.175	0.5 * V <sub>DDQ</sub> + 0.175	V	2

#### Note:

- 1.  $V_{ID}(AC)$  specifies the input differential voltage  $|V_{TR} V_{CP}|$  required for switching, where  $V_{TR}$  is the true input signal (such as CK, DQS, LDQS or UDQS) and  $V_{CP}$  is the complementary input signal (such as  $\overline{CK}$ ,  $\overline{DQS}$ ,  $\overline{LDQS}$  or  $\overline{UDQS}$ ). The minimum value is equal to  $V_{IH}$  (AC)  $V_{IL}(AC)$ .
- 2. The typical value of V<sub>IX</sub>(AC) is expected to be about 0.5 \* V<sub>DDQ</sub> of the transmitting device and V<sub>IX</sub>(AC) is expected to track variations in V<sub>DDQ</sub> . V<sub>IX</sub>(AC) indicates the voltage at which differential input signals must cross.
- For information related to V<sub>PEAK</sub> value, Refer to overshoot/undershoot specification in device operation and timing datasheet; maximum peak amplitude allowed for overshoot and undershoot.



#### < Differential signal levels >

#### 7.7 Differential AC output parameters

Symbol	Parameter	Min.	Max.	Units	Note
V <sub>OX</sub> (AC)	AC differential cross point voltage	0.5 * V <sub>DDQ</sub> - 0.125	0.5 * V <sub>DDQ</sub> + 0.125	V	1

#### Note:

1. The typical value of  $V_{OX}(AC)$  is expected to be about 0.5 \*  $V_{DDQ}$  of the transmitting device and  $V_{OX}(AC)$  is expected to track variations in  $V_{DDQ}$ .  $V_{OX}(AC)$  indicates the voltage at which differential output signals must cross.

#### 8.0 ODT DC electrical characteristics

PARAMETER/CONDITION	SYMBOL	MIN	MOM	MAX	UNITS	NOTES
Rtt effective impedance value for EMRS(A6,A2)=0,1; 75 ohm	Rtt1(eff)	60	86	103	ohm	1
Rtt effective impedance value for EMRS(A6,A2)=1,0; 150 ohm	Rtt2(eff)	120	173	207	ohm	1
Rtt effective impedance value for EMRS(A6,A2)=1,1; 50 ohm	Rtt3(eff)	40	58	67	ohm	1
Deviation of VM with respect to V <sub>DDQ</sub> /2	delta VM	- 6		+ 6	%	1

Note: Test condition for Rtt measurements

Measurement Definition for Rtt(eff): Apply  $V_{IH}$  (AC) and  $V_{IL}$  (AC) to test pin separately, then measure current  $I(V_{IH}$  (AC)) and  $I(V_{IL}$  (AC)) respectively.  $V_{IH}$  (AC),  $V_{IL}$  (AC), and  $V_{DDQ}$  values defined in SSTL\_18

$$Rtt(eff) = \frac{V_{IH} (AC) - V_{IL} (AC)}{I(V_{IH} (AC)) - I(V_{II} (AC))}$$

delta 
$$V_M = \begin{pmatrix} 2 \times V_M \\ V_{DDQ} \end{pmatrix} - 1 \times 100\%$$

Measurement Definition for V<sub>M</sub>: Measure voltage (V<sub>M</sub>) at test pin (midpoint) with no load.



# 9.0 IDD Specification Parameters and Test Conditions

(IDD values are for full operating range of Voltage and Temperature, Notes 1 - 5)

Symbol	Proposed Conditions		Units	Notes
IDD0	Operating one bank active-precharge current; tCK = tCK(IDD), tRC = tRC(IDD), tRAS = tRASmin(IDD); CKE is HIGH, CS is Address bus inputs are SWITCHING; Data bus inputs are SWITCHING	s HIGH between valid commands;	mA	
IDD1	Operating one bank active-read-precharge current; IOUT = 0mA; BL = 4, CL = $CL(IDD)$ , AL = 0; tCK = tCK(IDD), tRC = tRC (IDD) tRCD(IDD); CKE is HIGH, $\overline{CS}$ is HIGH between valid commands; Address bus is same as IDD4W		mA	
IDD2P	Precharge power-down current; All banks idle; tCK = tCK(IDD); CKE is LOW; Other control and address bus inpFLOATING	outs are STABLE; Data bus inputs are	mA	
IDD2Q	Precharge quiet standby current; All banks idle; tCK = tCK(IDD); CKE is HIGH, CS is HIGH; Other control and a bus inputs are FLOATING	address bus inputsare STABLE; Data	mA	
IDD2N	Precharge standby current; All banks idle; tCK = tCK(IDD); CKE is HIGH, CS is HIGH; Other control and a Data bus inputs are SWITCHING	mA		
	Active power-down current;	Fast PDN Exit MRS(12) = 0	mA	
IDD3P	All banks open; tCK = tCK(IDD); CKE is LOW; Other control and address bus inputs are STABLE; Data bus inputs are FLOATING	Slow PDN Exit MRS(12) = 1	mA	
IDD3N	Active standby current; All banks open; tCK = tCK(IDD), tRAS = tRASmax(IDD), tRP = tRP(IDD); CKI commands; Other control and address bus inputs are SWITCHING; Data bus		mA	
IDD4W	Operating burst write current; All banks open, Continuous burst writes; BL = 4, CL = CL(IDD), AL = 0; tCK = t = tRP(IDD); CKE is HIGH, CS is HIGH between valid commands; Address bu inputs are SWITCHING		mA	
IDD4R	Operating burst read current; All banks open, Continuous burst reads, IOUT = 0mA; BL = 4, CL = CL(IDD), $\mu$ max(IDD), tRP = tRP(IDD); CKE is HIGH, $\overline{\text{CS}}$ is HIGH between valid comman ING; Data pattern is same as IDD4W		mA	
IDD5B	Burst auto refresh current; tCK = tCK(IDD); Refresh command at every tRFC(IDD) interval; CKE is HIGH mands; Other control and address bus inputs are SWITCHING; Data bus input	mA		
	Self refresh current;	Normal	mA	
IDD6	CK and $\overline{\text{CK}}$ at 0V; CKE $\leq$ 0.2V; Other control and address bus inputs are FLOATING; Data bus inputs are FLOATING	Low Power	mA	
IDD7	Operating bank interleave read current;  All bank interleaving reads, IOUT = 0mA; BL = 4, CL = CL(IDD), AL = <sup>†</sup> RCD(ID = tRC(IDD), tRRD = tRRD(IDD), tFAW = tFAW(IDD), tRCD = 1*tCK(IDD); CKE commands; Address bus inputs are STABLE during DESELECTs; Data patter lowing page for detailed timing conditions	E is HIGH, CS is HIGH between valid	mA	



#### Note:

- 1. IDD specifications are tested after the device is properly initialized
- 2. Input slew rate is specified by AC Parametric Test Condition
- 3. IDD parameters are specified with ODT disabled.
- 4. Data bus consists of DQ, DM, DQS,  $\overline{DQS}$ , RDQS,  $\overline{RDQS}$ , LDQS,  $\overline{LDQS}$ , UDQS, and  $\overline{UDQS}$ . IDD values must be met with all combinations of EMRS bits 10 and 11.
- 5. Definitions for IDD

LOW is defined as  $V_{IN} \le V_{IL}AC(max)$ 

HIGH is defined as  $V_{IN} \geq V_{IH} AC (min)$ 

STABLE is defined as inputs stable at a HIGH or LOW level

FLOATING is defined as inputs at  $V_{REF} = V_{DDQ}/2$ 

SWITCHING is defined as:

inputs changing between HIGH and LOW every other clock cycle (once per two clocks) for address and control signals, and

inputs changing between HIGH and LOW every other data transfer (once per clock) for DQ signals not including masks or strobes.

For purposes of IDD testing, the following parameters are utilized

	DDR2-667	Haita
Parameter	5-5-5	Units
CL(IDD)	5	tCK
tRCD(IDD)	15	ns
tRC(IDD)	60	ns
tRRD(IDD)-x4/x8	7.5	ns
tRRD(IDD)-x16	10	ns
tCK(IDD)	3	ns
tRASmin(IDD)	45	ns
tRP(IDD)	15	ns
tRFC(IDD)	127.5	ns

#### Detailed IDD7

The detailed timings are shown below for IDD7.

Legend: A = Active; RA = Read with Autoprecharge; D = Deselect

#### IDD7: Operating Current: All Bank Interleave Read operation

All banks are being interleaved at minimum tRC(IDD) without violating tRRD(IDD) and tFAW(IDD) using a burst length of 4. Control and address bus inputs are STABLE during DESELECTs. IOUT = 0mA

#### Timing Patterns for 8bank devices x4/ x8

- -DDR2-667 5/5/5 : A0 RA0 D A1 RA1 D A2 RA2 D A3 RA3 D D A4 RA4 D A5 RA5 D A6 RA6 D A7 RA7 D D
- -DDR2-800 6/6/6 : A0 RA0 D A1 RA1 D A2 RA2 D A3 RA3 D D D A4 RA4 D A5 RA5 D A6 RA6 D A7 RA7 D D D
- -DDR2-800 5/5/5 : A0 RA0 D A1 RA1 D A2 RA2 D A3 RA3 D D D A4 RA4 D A5 RA5 D A6 RA6 D A7 RA7 D D D

# 10.0 DDR2 SDRAM IDD Spec Table

	256Mx4 (K4T1G044QQ)		
Symbol	667@CL=5	Unit	Notes
	YE6		
IDD0	65	mA	
IDD1	75	mA	
IDD2P	15	mA	
IDD2Q	30	mA	
IDD2N	35	mA	
IDD3P-F	35	mA	
IDD3P-S	18	mA	
IDD3N	45	mA	
IDD4W	100	mA	
IDD4R	110	mA	
IDD5	130	mA	
IDD6	15	mA	
IDD7	220	mA	

Symbol	128Mx8 (K4T1G084QQ) 667@CL=5 YE6	Unit	Notes
IDD0	65	mA	
IDD1	75	mA	
IDD2P	15	mA	
IDD2Q	30	mA	
IDD2N	35	mA	
IDD3P-F	35	mA	
IDD3P-S	18	mA	
IDD3N	45	mA	
IDD4W	100	mA	
IDD4R	115	mA	
IDD5	130	mA	
IDD6	15	mA	
IDD7	225	mA	

# 11.0 Input/Output capacitance

Parameter	Symbol	DDR	Units	
Faranteter	Syllibol	Min	Max	Offics
Input capacitance, CK and CK	CCK	1.0	2.0	pF
Input capacitance delta, CK and CK	CDCK	х	0.25	pF
Input capacitance, all other input-only pins	CI	1.0	2.0	pF
Input capacitance delta, all other input-only pins	CDI	х	0.25	pF
Input/output capacitance, DQ, DM, DQS, DQS	CIO	2.5	3.5	pF
Input/output capacitance delta, DQ, DM, DQS, DQS	CDIO	х	0.5	pF

# 12.0 Electrical Characteristics & AC Timing for DDR2-667

 $(0 \text{ °C} \le T_{\text{OPER}} \le 95 \text{ °C}; V_{\text{DDQ}} = 1.55 \text{V} + 0.05 \text{V}; V_{\text{DD}} = 1.55 \text{V} + 0.05 \text{V})$ 

# 12.1 Refresh Parameters by Device Density

Parameter		Symbol	256Mb	512Mb	1Gb	2Gb	4Gb	Units
Refresh to active/Refresh command time	tRFC		75	105	127.5	195	327.5	ns
Average periodic refresh interval	tREFI	$0  ^{\circ}\text{C} \le \text{T}_{\text{CASE}} \le 85 ^{\circ}\text{C}$	7.8	7.8	7.8	7.8	7.8	μS
Average periodic refresh interval		$85  ^{\circ}\text{C} < \text{T}_{CASE} \leq 95  ^{\circ}\text{C}$	3.9	3.9	3.9	3.9	3.9	μS

# 12.2 Speed Bins and CL, tRCD, tRP, tRC and tRAS for Corresponding Bin

Speed	DDR2-	667(E6)		
Bin (CL - tRCD - tRP)	5 -	Units		
Parameter	min	max		
tCK, CL=3	5	8	ns	
tCK, CL=4	3.75	8	ns	
tCK, CL=5	3	8	ns	
tCK, CL=6	-	-	ns	
tRCD	15	-	ns	
tRP	15	-	ns	
tRC	60	-	ns	
tRAS	45	70000	ns	

# 12.3 Timing Parameters by Speed Grade

(For information related to the entries in this table, refer to both the general notes and the specific notes following this table.)

Downwater	Counch of	DDR	2-667	Unita	Notes
Parameter	Symbol	min	max	Units	Notes
DQ output access time from CK/CK	tAC	-450	450	ps	40
DQS output access time from CK/CK	tDQSCK	-400	400	ps	40
Average clock HIGH pulse width	tCH(avg)	0.48	0.52	tCK(avg)	35,36
Average clock LOW pulse width	tCL(avg)	0.48	0.52	tCK(avg)	35,36
CK half pulse period	tHP	Min(tCL(abs), tCH(abs))	х	ps	37
Average clock period	tCK(avg)	3000	8000	ps	35,36
DQ and DM input hold time	tDH(base)	175	х	ps	6,7,8,21,28,31
DQ and DM input setup time	tDS(base)	100	х	ps	6,7,8,20,28,31
Control & Address input pulse width for each input	tIPW	0.6	х	tCK(avg)	
DQ and DM input pulse width for each input	tDIPW	0.35	х	tCK(avg)	
Data-out high-impedance time from CK/CK	tHZ	х	tAC(max)	ps	18,40
DQS/DQS low-impedance time from CK/CK	tLZ(DQS)	tAC(min)	tAC(max)	ps	18,40
DQ low-impedance time from CK/CK	tLZ(DQ)	2* tAC(min)	tAC(max)	ps	18,40
DQS-DQ skew for DQS and associated DQ signals	tDQSQ	х	240	ps	13
DQ hold skew factor	tQHS	х	340	ps	38
DQ/DQS output hold time from DQS	tQH	tHP - tQHS	х	ps	39
DQS latching rising transitions to associated clock edges	tDQSS	-0.25	0.25	tCK(avg)	30
DQS input HIGH pulse width	tDQSH	0.35	х	tCK(avg)	
DQS input LOW pulse width	tDQSL	0.35	х	tCK(avg)	
DQS falling edge to CK setup time	tDSS	0.2	х	tCK(avg)	30
DQS falling edge hold time from CK	tDSH	0.2	х	tCK(avg)	30
Mode register set command cycle time	tMRD	2	х	nCK	
MRS command to ODT update delay	tMOD	0	12	ns	32
Write postamble	tWPST	0.4	0.6	tCK(avg)	10
Write preamble	tWPRE	0.35	х	tCK(avg)	
Address and control input hold time	tlH(base)	275	х	ps	5,7,9,23,29
Address and control input setup time	tlS(base)	200	х	ps	5,7,9,22,29
Read preamble	tRPRE	0.9	1.1	tCK(avg)	19,41
Read postamble	tRPST	0.4	0.6	tCK(avg)	19,42
Activate to activate command period for 1KB page size products	tRRD	7.5	х	ns	4,32
Activate to activate command period for 2KB page size products	tRRD	10	х	ns	4,32



Daniel de la constante de la c	O	DDR	2-667	1114	Notes
Parameter	Symbol	min	max	Units	Notes
Four Activate Window for 1KB page size products	tFAW	37.5	х	ns	32
Four Activate Window for 2KB page size products	tFAW	50	х	ns	32
CAS to CAS command delay	tCCD	2	х	nCK	
Write recovery time	tWR	15	х	ns	32
Auto precharge write recovery + precharge time	tDAL	WR + tnRP	х	nCK	33
Internal write to read command delay	tWTR	7.5	х	ns	24,32
Internal read to precharge command delay	tRTP	7.5	х	ns	3,32
Exit self refresh to a non-read command	tXSNR	tRFC + 10	х	ns	32
Exit self refresh to a read command	tXSRD	200	х	nCK	
Exit precharge power down to any command	tXP	2	х	nCK	
Exit active power down to read command	tXARD	2	х	nCK	1
Exit active power down to read command (slow exit, lower power)	tXARDS	7 - AL	х	nCK	1,2
CKE minimum pulse width (HIGH and LOW pulse width)	tCKE	3	х	nCK	27
ODT turn-on delay	tAOND	2	2	nCK	16
ODT turn-on	tAON	tAC(min)	tAC(max)+0.7	ns	6,16,40
ODT turn-on (Power-Down mode)	tAONPD	tAC(min)+2	2*tCK(avg) +tAC(max)+1	ns	
ODT turn-off delay	tAOFD	2.5	2.5	nCK	17,45
ODT turn-off	tAOF	tAC(min)	tAC(max)+0.6	ns	17,43,45
ODT turn-off (Power-Down mode)	tAOFPD	tAC(min)+2	2.5*tCK(avg) +tAC(max)+1	ns	
ODT to power down entry latency	tANPD	3	х	nCK	
ODT power down exit latency	tAXPD	8	х	nCK	
OCD drive mode output delay	tOIT	0	12	ns	32
Minimum time clocks remains ON after CKE asynchronously drops LOW	tDelay	tIS+tCK(avg) +tIH	х	ns	15



# 13.0 General notes, which may apply for all AC parameters

#### 1. DDR2 SDRAM AC timing reference load

Figure 1 represents the timing reference load used in defining the relevant timing parameters of the part. It is not intended to be either a precise repre sentation of the typical system environment or a depiction of the actual load presented by a production tester. System designers will use IBIS or other simulation tools to correlate the timing reference load to a system environment. Manufacturers will correlate to their production test conditions (generally a coaxial transmission line terminated at the tester electronics).

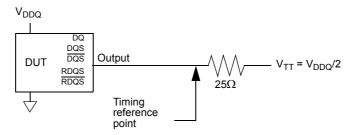


Figure 1 - AC Timing Reference Load

The output timing reference voltage level for single ended signals is the crosspoint with  $V_{TT}$ . The output timing reference voltage level for differential signals is the crosspoint of the true (e.g. DQS) and the complement (e.g.  $\overline{DQS}$ ) signal.

#### 2. Slew Rate Measurement Levels

- a) Output slew rate for falling and rising edges is measured between V<sub>TT</sub> 250 mV and V<sub>TT</sub> + 250 mV for single ended signals. For differential signals (e.g. DQS DQS) output slew rate is measured between DQS DQS = 500 mV and DQS DQS = + 500 mV. Output slew rate is guaranteed by design, but is not necessarily tested on each device.
- b) Input slew rate for single ended signals is measured from  $V_{REF}(DC)$  to  $V_{IH}(AC)$ ,min for rising edges and from  $V_{REF}(DC)$  to  $V_{IL}(AC)$ ,max for falling edges. For differential signals (e.g. CK  $\overline{CK}$ ) slew rate for rising edges is measured from CK  $\overline{CK}$  = 250 mV to CK  $\overline{CK}$  = + 500 mV (+ 250 mV to 500 mV for falling edges).
- c) V<sub>ID</sub> is the magnitude of the difference between the input voltage on CK and the input voltage on CK, or between DQS and DQS for differential strobe.

#### 3. DDR2 SDRAM output slew rate test load

Output slew rate is characterized under the test conditions as shown in Figure 2.

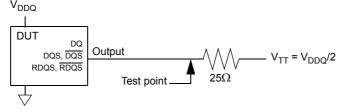


Figure 2 - Slew Rate Test Load

#### 4. Differential data strobe

DDR2 SDRAM pin timings are specified for either single ended mode or differential mode depending on the setting of the EMRS "Enable DQS" mode bit; timing advantages of differential mode are realized in system design. The method by which the DDR2 SDRAM pin timings are measured is mode dependent. In single ended mode, timing relationships are measured relative to the rising or falling edges of DQS crossing at  $V_{REF}$ . In differential mode, these timing relationships are measured relative to the crosspoint of DQS and its complement,  $\overline{DQS}$ . This distinction in timing methods is guaranteed by design and characterization. Note that when differential data strobe mode is disabled via the EMRS, the complementary pin,  $\overline{DQS}$ , must be tied externally to  $V_{SS}$  through a 20  $\Omega$  to 10 k $\Omega$  resistor to insure proper operation.

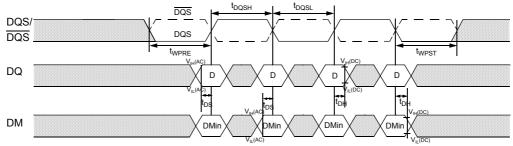


Figure 3 - Data Input (Write) Timing

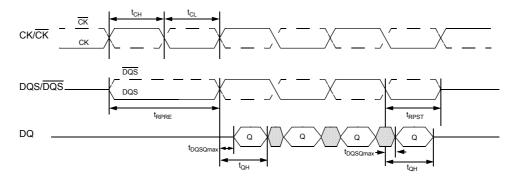


Figure 4 - Data Output (Read) Timing

- 5. AC timings are for linear signal transitions. See Specific Notes on derating for other signal transitions.
- **6.** All voltages are referenced to V<sub>SS</sub>.
- 7. These parameters guarantee device behavior, but they are not necessarily tested on each device. They may be guaranteed by device design or tester correlation.
- 8. Tests for AC timing, IDD, and electrical (AC and DC) characteristics, may be conducted at nominal reference/supply voltage levels, but the related specifications and device operation are guaranteed for the full voltage range specified.

# 14.0 Specific Notes for dedicated AC parameters

- 1. User can choose which active power down exit timing to use via MRS (bit 12). tXARD is expected to be used for fast active power down exit timing. tXARDS is expected to be used for slow active power down exit timing.
- 2. AL = Additive Latency.
- 3. This is a minimum requirement. Minimum read to precharge timing is AL + BL / 2 provided that the tRTP and tRAS(min) have been satisfied.
- 4. A minimum of two clocks (2 x tCK or 2 x nCK) is required irrespective of operating frequency.
- 5. Timings are specified with command/address input slew rate of 1.0 V/ns.
- 6. Timings are specified with DQs, DM, and DQS's (DQS/RDQS in single ended mode) input slew rate of 1.0V/ns.
- 7. Timings are specified with CK/CK differential slew rate of 2.0 V/ns. Timings are guaranteed for DQS signals with a differential slew rate of 2.0 V/ns in differential strobe mode and a slew rate of 1.0 V/ns in single ended mode.
- 8. Data setup and hold time derating.

Table 1 - DDR2-400/533 tDS/tDH derating with differential data strobe

		Δ	tDS, ∆tl	DH Der	ating Va	alues o	f DDR2	-400, DI	DR2-53	3 (ALL	units in	'ps', th	e note	applies	to enti	re Table	e)		
								D	QS,DQ	S Differ	ential S	lew Ra	te						
		4.0 \	V/ns	3.0	V/ns	2.0	V/ns	1.8	V/ns	1.6	V/ns	1.4\	//ns	1.2\	//ns	1.0\	//ns	0.8	//ns
		∆tDS	∆tDH	∆tDS	∆tDH	∆tDS	∆tDH	∆tDS	∆tDH	∆tDS	∆tDH	∆tDS	∆tDH	∆tDS	∆tDH	∆tDS	∆tDH	∆tDS	∆tDH
	2.0	125	45	125	45	125	45	-	-	-	-	-	-	-	-	-	-	-	-
	1.5	83	21	83	21	83	21	95	33	-	-	-	-	-	-	-	-	-	-
	1.0	0	0	0	0	0	0	12	12	24	24	-	-	-	-	-	-	-	-
DQ	0.9	-	-	-11	-14	-11	-14	1	-2	13	10	25	22	-	-	-	-	-	-
Siew	8.0	-	-	-	-	-25	-31	-13	-19	-1	-7	11	5	23	17	-	-	-	-
V/ns	0.7	-	-	-	-	-	-	-31	-42	-19	-30	-7	-18	5	-6	17	6	-	-
	0.6	-	-	-	-	-	-	-	-	-43	-59	-31	-47	-19	-35	-7	-23	5	-11
	0.5	-	•	-	-	•	-	-	-	-	•	-74	-89	-62	-77	-50	-65	-38	-53
	0.4	-	-	-	-	-	-	-	-	-	-	-	-	-127	-140	-115	-128	-103	-116

Table 2 - DDR2-667/800 tDS/tDH derating with differential data strobe

		Δt	DS, ∆tD	)H Dera	ting Va	lues fo	r DDR2	-667, D	DR2-80	0 (ALL	units ir	ı 'ps', tl	ne note	applies	s to ent	ire Tab	le)		
								D	QS,DQ	S Differ	ential S	lew Ra	te						
		4.0	V/ns	3.0	V/ns	2.0	V/ns	1.8	V/ns	1.6	V/ns	1.4\	//ns	1.2\	//ns	1.0\	//ns	0.8	V/ns
		∆tDS	∆tDH	∆tDS	∆tDH	∆tDS	∆tDH	∆tDS	∆tDH	∆tDS	∆tDH	∆tDS	∆tDH	∆tDS	∆tDH	∆tDS	∆tDH	∆tDS	∆tDH
	2.0	100	45	100	45	100	45	-	-	-	-	-	-	-	-	-	-	-	-
	1.5	67	21	67	21	67	21	79	33	-	-	-	-	-	-	-	-	-	-
	1.0	0	0	0	0	0	0	12	12	24	24	-	-	-	-	-	-	-	-
DQ	0.9	-	-	-5	-14	-5	-14	7	-2	19	10	31	22	•	-	•	-	-	-
Slew	8.0	-	-	-	-	-13	-31	-1	-19	11	-7	23	5	35	17	-	-	-	-
V/ns	0.7	-	-	-	-	-	-	-10	-42	2	-30	14	-18	26	-6	38	6	-	-
	0.6	-	-	-	-	•	-	-	-	-10	-59	2	-47	14	-35	26	-23	38	-11
	0.5	-	-	-	-	-	-	-	-	-	-	-24	-89	-12	-77	0	-65	12	-53
	0.4	-	-	-	-	-	-	-	-	-	-	-	-	-52	-140	-40	-128	-28	-116



Table 3 - DDR2-400/533 tDS1/tDH1 derating with single-ended data strobe

		∆tD	S1, ∆tD	H1 Der	ating V	alues fo	or DDR	2-400, E	DR2-5	33(All u	nits in '	ps'; the	e note a	pplies	to the e	ntire ta	ble)		
									DQS Si	ngle-en	ded Sle	ew Rate	)						
		2.0 V/ns 1.5 V/ns 1.0 V/ns 0.9 V/ns 0.8 V/ns 0.7 V/n										V/ns 0.6 V/ns		V/ns	0.5 V/ns		0.4	V/ns	
		∆tDS 1	∆tDH 1	∆tDS 1	∆tDH 1	∆tDS 1	∆tDH 1	∆tDS 1	∆tDH 1	∆tDS 1	∆tDH 1	∆tDS 1	∆tDH 1	∆tDS 1	∆tDH 1	∆tDS 1	∆tDH 1	∆tDS 1	∆tDH 1
	2.0	188	188	167	146	125	63	-	-	-	-	-	-	-	-	-	-	-	-
	1.5	146	167	125	125	83	42	81	43	-	•	-	-	-	•	-	-	-	-
	1.0	63	125	42	83	0	0	-2	1	-7	-13	-	-	-	•	-	-	-	-
DQ	0.9	-	-	31	69	-11	-14	-13	-13	-18	-27	-29	-45	-	-	-	-	-	-
Slew	8.0	-	-	-	-	-25	-31	-27	-30	-32	-44	-43	-62	-60	-86	-	-	-	-
V/ns	0.7	-	-	-	-	-	-	-45	-53	-50	-67	-61	-85	-78	-109	-108	-152	-	-
	0.6	-	-	-	-	-	-	-	-	-74	-96	-85	-114	-102	-138	-138	-181	-183	-246
	0.5	-	-	-	-	-	-	-	-	-	-	-128	-156	-145	-180	-175	-223	-226	-288
	0.4	-	-	-	-	-	-	-	-	-	-	-	-	-210	-243	-240	-286	-291	-351

For all input signals the total tDS (setup time) and tDH (hold time) required is calculated by adding the data sheet tDS(base) and tDH(base) value to the  $\Delta$ tDS and  $\Delta$ tDH derating value respectively. Example: tDS (total setup time) =tDS(base) + $\Delta$ tDS.

Setup (tDS) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of  $V_{REF}(DC)$  and the first crossing of  $V_{IH}(AC)$ min. Setup (tDS) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of  $V_{REF}(DC)$  and the first crossing of  $V_{IL}(AC)$ max. If the actual signal is always earlier than the nominal slew rate line between shaded ' $V_{REF}(DC)$ ' to ac region', use nominal slew rate for derating value (See Figure 5 for differential data strobe and Figure 6 for single-ended data strobe.) If the actual signal is later than the nominal slew rate line anywhere between shaded ' $V_{REF}(DC)$ ' to ac region', the slew rate of a tangent line to the actual signal from the ac level to dc level is used for derating value (see Figure 7 for differential data strobe and Figure 8 for single-ended data strobe)

Hold (tDH) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of  $V_{IL}(DC)$ max and the first crossing of  $V_{REF}(DC)$ . Hold (tDH) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of  $V_{IH}(DC)$ min and the first crossing of  $V_{REF}(DC)$ . If the actual signal is always later than the nominal slew rate line between shaded 'dc level to  $V_{REF}(DC)$  region', use nominal slew rate for derating value (see Figure 9 for differential data strobe and Figure 10 for single-ended data strobe) If the actual signal is earlier than the nominal slew rate line anywhere between shaded 'dc to  $V_{REF}(DC)$  region', the slew rate of a tangent line to the actual signal from the dc level to  $V_{REF}(DC)$  level is used for derating value (see Figure 11 for differential data strobe and Figure 12 for single-ended data strobe)

Although for slow slew rates the total setup time might be negative (i.e. a valid input signal will not have reached  $V_{IH/IL}(AC)$  at the time of the rising clock transition) a valid input signal is still required to complete the transition and reach  $V_{IH/IL}(AC)$ .

For slew rates in between the values listed in Tables 1, 2 and 3, the derating values may obtained by linear interpolation.

These values are typically not subject to production test. They are verified by design and characterization.



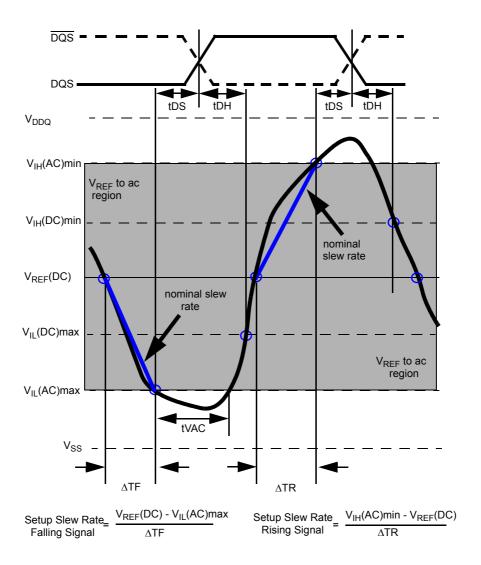
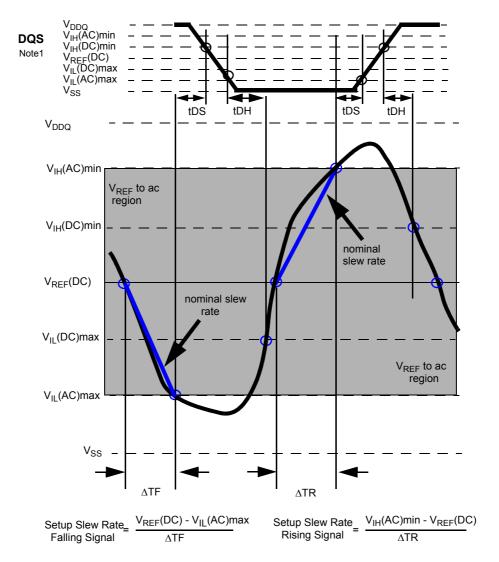


Figure 5 - Illustration of nominal slew rate for tDS (differential DQS, DQS)



Note : DQS signal must be monotonic between  $V_{IL}(AC)max$  and  $V_{IH}(AC)min$ .

Figure 6 - Illustration of nominal slew rate for tDS (single-ended DQS)

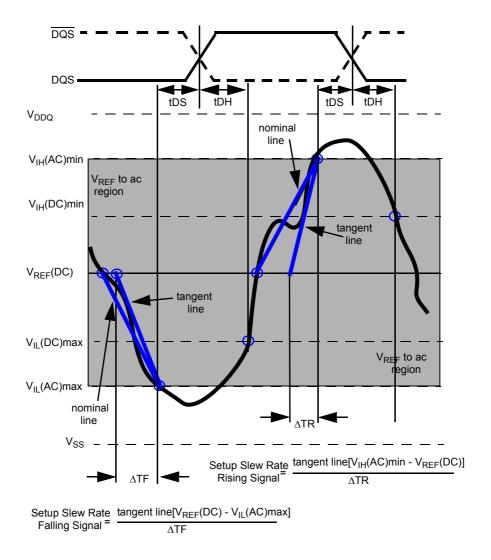
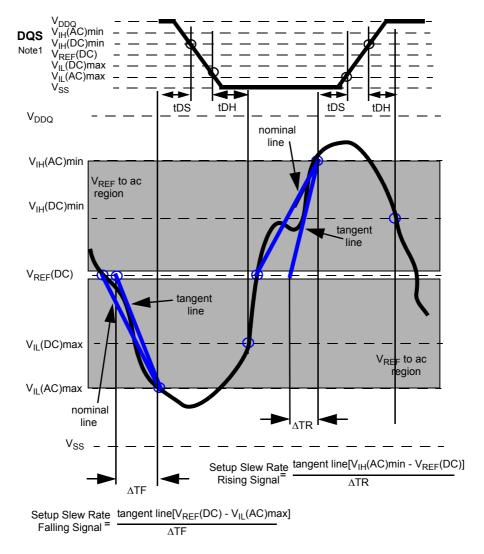


Figure 7 - Illustration of tangent line for tDS (differential DQS, DQS)



Note : DQS signal must be monotonic between  $V_{IL}(DC)$ max and  $V_{IH}(DC)$ min.

Figure 8 - Illustration of tangent line for tDS (single-ended DQS)

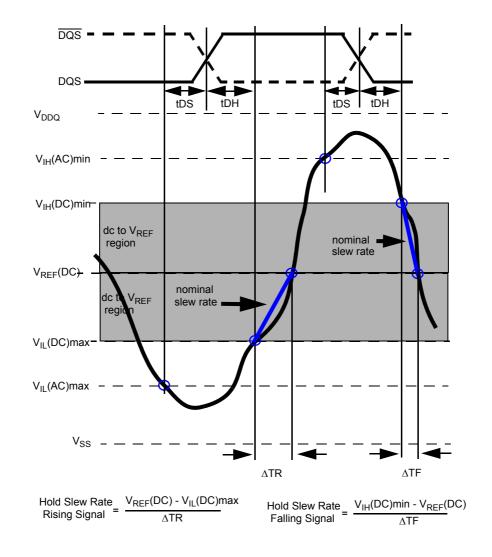
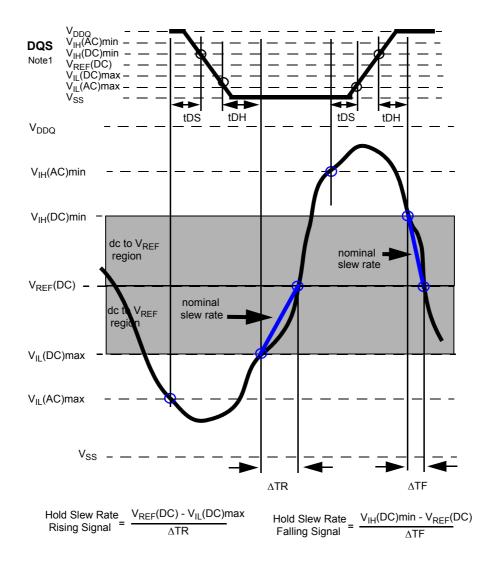


Figure 9 - Illustration of nominal slew rate for tDH (differential DQS, DQS)



Note : DQS signal must be monotonic between  $V_{IL}(DC)$ max and  $V_{IH}(DC)$ min.

Figure 10 - Illustration of nominal slew rate for tDH (single-ended DQS)

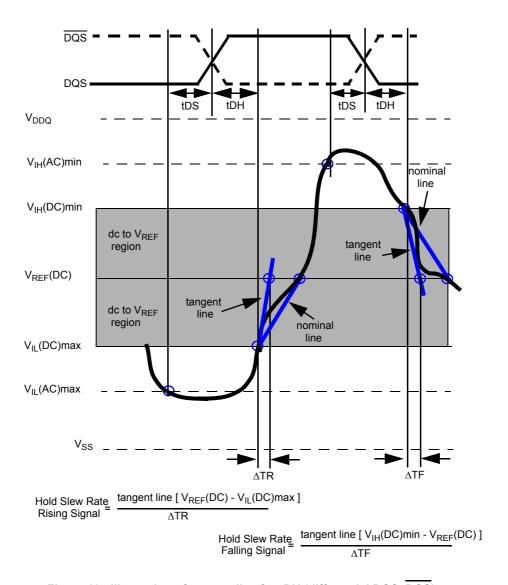
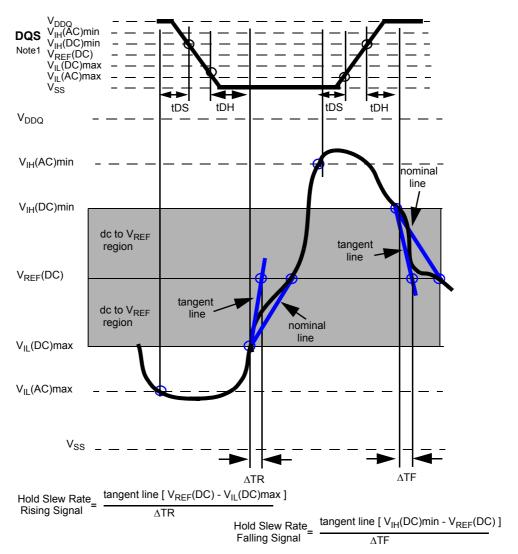


Figure 11 - Illustration of tangent line for tDH (differential DQS, DQS)



Note : DQS signal must be monotonic between  $V_{IL}(DC)$ max and  $V_{IH}(DC)$ min.

Figure 12 - Illustration of tangent line for tDH (single-ended DQS)

9. tIS and tIH (input setup and hold) derating

Table 4 - Derating values for DDR2-400, DDR2-533

∆tIS, ∆tIH Derating Values for DDR2-400, DDR2-533											
		CK, CK Differential Slew Rate									
		2.0 \	V/ns	1.5 V/ns		1.0 V/ns					
		ΔtIS	∆tIH	ΔtIS	∆tIH	ΔtIS	∆tIH	Units	Notes		
	4.0	+187	+94	+217	+124	+247	+154	ps	1		
	3.5	+179	+89	+209	+119	+239	+149	ps	1		
	3.0	+167	+83	+197	+113	+227	+143	ps	1		
	2.5	+150	+75	+180	+105	+210	+135	ps	1		
	2.0	+125	+45	+155	+75	+185	+105	ps	1		
	1.5	+83	+21	+113	+51	+143	+81	ps	1		
	1.0	0	0	+30	+30	+60	+60	ps	1		
	0.9	-11	-14	+19	+16	+49	+46	ps	1		
Command/ Address Slew	0.8	-25	-31	+5	-1	+35	+29	ps	1		
rate(V/ns)	0.7	-43	-54	-13	-24	+17	+6	ps	1		
	0.6	-67	-83	-37	-53	-7	-23	ps	1		
	0.5	-110	-125	-80	-95	-50	-65	ps	1		
	0.4	-175	-188	-145	-158	-115	-128	ps	1		
	0.3	-285	-292	-255	-262	-225	-232	ps	1		
	0.25	-350	-375	-320	-345	-290	-315	ps	1		
	0.2	-525	-500	-495	-470	-465	-440	ps	1		
	0.15	-800	-708	-770	-678	-740	-648	ps	1		
	0.1	-1450	-1125	-1420	-1095	-1390	-1065	ps	1		



Table 5 - Derating values for DDR2-667, DDR2-800

∆tIS and ∆tIH Derating Values for DDR2-667, DDR2-800											
	CK, CK Differential Slew Rate										
	2.0 V/ns			1.5 V/ns		1.0 V/ns		1114	Neter		
		ΔtIS	∆tIH	∆tIS	∆tIH	ΔtIS	∆tIH	Units	Notes		
	4.0	+150	+94	+180	+124	+210	+154	ps	1		
	3.5	+143	+89	+173	+119	+203	+149	ps	1		
	3.0	+133	+83	+163	+113	+193	+143	ps	1		
	2.5	+120	+75	+150	+105	+180	+135	ps	1		
	2.0	+100	+45	+130	+75	+160	+105	ps	1		
	1.5	+67	+21	+97	+51	+127	+81	ps	1		
	1.0	0	0	+30	+30	+60	+60	ps	1		
	0.9	-5	-14	+25	+16	+55	+46	ps	1		
Command/ Address Slew	0.8	-13	-31	+17	-1	+47	+29	ps	1		
rate(V/ns)	0.7	-22	-54	+8	-24	+38	+6	ps	1		
, ,	0.6	-34	-83	-4	-53	+26	-23	ps	1		
	0.5	-60	-125	-30	-95	0	-65	ps	1		
	0.4	-100	-188	-70	-158	-40	-128	ps	1		
	0.3	-168	-292	-138	-262	-108	-232	ps	1		
	0.25	-200	-375	-170	-345	-140	-315	ps	1		
	0.2	-325	-500	-295	-470	-265	-440	ps	1		
	0.15	-517	-708	-487	-678	-457	-648	ps	1		
	0.1	-1000	-1125	-970	-1095	-940	-1065	ps	1		

For all input signals the total tIS (setup time) and tIH (hold time) required is calculated by adding the data sheet tIS(base) and tIH(base) value to the  $\Delta$ tIS and  $\Delta$ tIH derating value respectively. Example: tIS (total setup time) = tIS(base) +  $\Delta$ tIS

Setup (tlS) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of  $V_{REF}(DC)$  and the first crossing of  $V_{IL}(AC)$ min. Setup (tlS) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of  $V_{REF}(DC)$  and the first crossing of  $V_{IL}(AC)$ max. If the actual signal is always earlier than the nominal slew rate line between shaded  $V_{REF}(DC)$  to ac region', use nominal slew rate for derating value (see Figure 13). If the actual signal is later than the nominal slew rate line anywhere between shaded  $V_{REF}(DC)$  to ac region', the slew rate of a tangent line to the actual signal from the ac level to dc level is used for derating value (see Figure 14).

Hold (tIH) nominal slew rate for a rising signal is defined as the slew rate between the last crossing of  $V_{IL}(DC)$ max and the first crossing of  $V_{REF}(DC)$ . Hold (tIH) nominal slew rate for a falling signal is defined as the slew rate between the last crossing of  $V_{IH}(DC)$ min and the first crossing of  $V_{REF}(DC)$ . If the actual signal is always later than the nominal slewrate line between shaded 'dc to  $V_{REF}(DC)$  region', use nominal slew rate for derating value (see Figure 15). If the actual signal is earlier than the nominal slew rate line anywhere between shaded 'dc to  $V_{REF}(DC)$  region', the slew rate of a tangent line to the actual signal from the dc level to  $V_{REF}(DC)$  level is used for derating value (see Figure 16).

Although for slow slew rates the total setup time might be negative (i.e. a valid input signal will not have reached  $V_{IH/IL}(AC)$  at the time of the rising clock transition) a valid input signal is still required to complete the transition and reach  $V_{IH/IL}(AC)$ .

For slew rates in between the values listed in Tables 4 and 5, the derating values may obtained by linear interpolation.

These values are typically not subject to production test. They are verified by design and characterization.



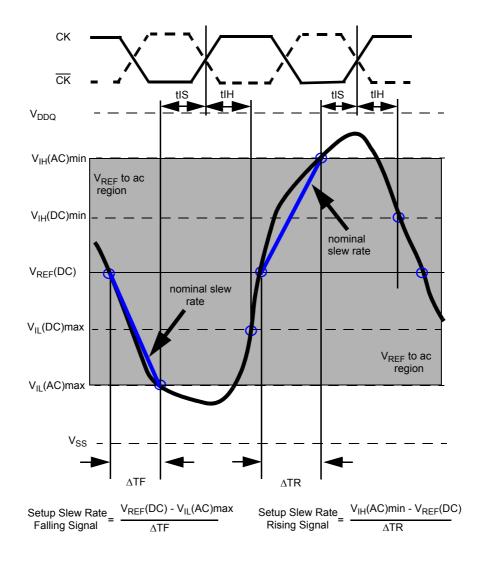


Figure 13 - Illustration of nominal slew rate for tIS

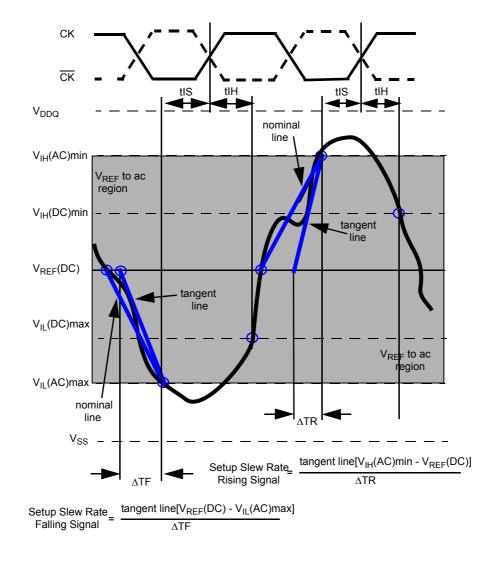


Figure 14 - Illustration of tangent line for tIS

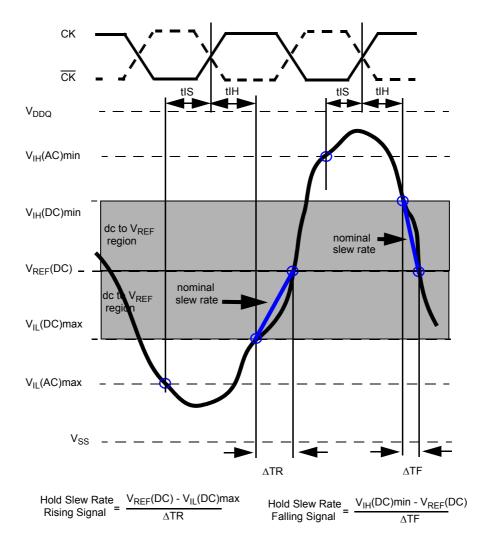


Figure 15 - Illustration of nominal slew rate for tlH

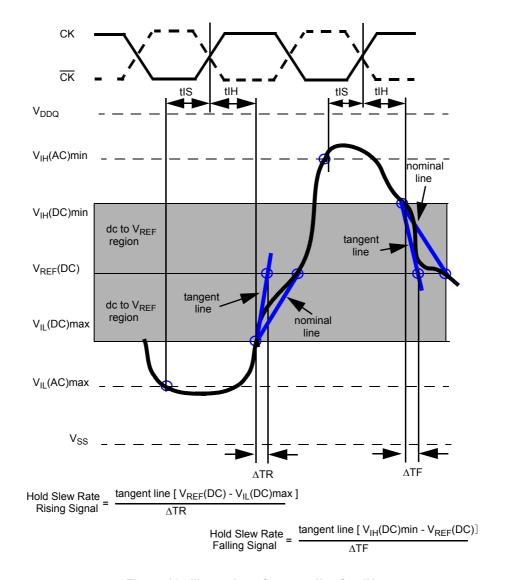


Figure 16 - Illustration of tangent line for tlH

- **10.** The maximum limit for this parameter is not a device limit. The device will operate with a greater value for this parameter, but system performance (bus turnaround) will degrade accordingly.
- 11. MIN (tCL, tCH) refers to the smaller of the actual clock LOW time and the actual clock HIGH time as provided to the device (i.e. this value can be greater than the minimum specification limits for tCL and tCH). For example, tCL and tCH are = 50% of the period, less the half period jitter (tJIT(HP)) of the clock source, and less the half period jitter due to crosstalk (tJIT(crosstalk)) into the clock traces.
- 12. tQH = tHP tQHS, where :

tHP = minimum half clock period for any given cycle and is defined by clock HIGH or clock LOW (tCH, tCL). tQHS accounts for:

- 1) The pulse duration distortion of on-chip clock circuits; and
- 2) The worst case push-out of DQS on one transition followed by the worst case pull-in of DQ on the next transition, both of which are, separately, due to data pin skew and output pattern effects, and p-channel to n-channel variation of the output drivers.
- 13. tDQSQ: Consists of data pin skew and output pattern effects, and p-channel to n-channel variation of the output drivers as well as output slew rate mismatch between DQS/ DQS and associated DQ in any given cycle.
- 14. tDAL = WR + RU{ tRP[ns] / tCK[ns] }, where RU stands for round up.

WR refers to the tWR parameter stored in the MRS. For tRP, if the result of the division is not already an integer, round up to the next highest integer. tCK refers to the application clock period.

Example: For DDR533 at tCK = 3.75ns with WR programmed to 4 clocks.

tDAL = 4 + (15 ns / 3.75 ns) clocks = 4 + (4) clocks = 8 clocks.

- 15. The clock frequency is allowed to change during self refresh mode or precharge power-down mode.
- **16.** ODT turn on time min is when the device leaves high impedance and ODT resistance begins to turn on. ODT turn on time max is when the ODT resistance is fully on. Both are measured from tAOND, which is interpreted differently per speed bin. For DDR2-400/533, tAOND is 10 ns (= 2 x 5 ns) after the clock edge that registered a first ODT HIGH if tCK = 5 ns. For DDR2-667/800, tAOND is 2 clock cycles after the clock edge that registered a first ODT HIGH counting the actual input clock edges.
- 17. ODT turn off time min is when the device starts to turn off ODT resistance. ODT turn off time max is when the bus is in high impedance. Both are measured from tAOFD, which is interpreted differently per speed bin. For DDR2-400/533, tAOFD is 12.5 ns (= 2.5 x 5 ns) after the clock edge that registered a first ODT LOW if tCK = 5 ns. For DDR2-667/800, if tCK(avg) = 3 ns is assumed, tAOFD is 1.5 ns (= 0.5 x 3 ns) after the second trailing clock edge counting from the clock edge that registered a first ODT LOW and by counting the actual input clock edges.
- 18. tHZ and tLZ transitions occur in the same access time as valid data transitions. These parameters are referenced to a specific voltage level which specifies when the device output is no longer driving (tHZ), or begins driving (tLZ). Figure 17 shows a method to calculate the point when device is no longer driving (tHZ), or beginsdriving (tLZ) by measuring the signal at two different voltages. The actual voltage measurement points are not critical as long as the calculation is consistent. tLZ(DQ) refers to tLZ of the DQS and tLZ(DQS) refers to tLZ of the (U/L/R)DQS and (U/L/R)DQS each treated as single-ended signal.
- 19. tRPST end point and tRPRE begin point are not referenced to a specific voltage level but specify when the device output is no longer driving (tRPST), or begins driving (tRPRE). Figure 17 shows a method to calculate these points when the device is no longer driving (tRPST), or begins driving (tRPRE) by measuring the signal at two different voltages. The actual voltage measurement points are not critical as long as the calculation is consistent.

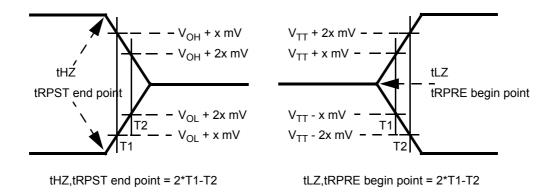


Figure 17 - Method for calculating transitions and endpoints



- DDR2 1.55V SDRAM
- 20. Input waveform timing tDS with differential data strobe enabled MR[bit10]=0, is referenced from the input signal crossing at the  $V_{IH}(AC)$  level to the differential data strobe crosspoint for a rising signal, and from the input signal crossing at the  $V_{IL}(AC)$  level to the differential data strobe crosspoint for a falling signal applied to the device under test. DQS,  $\overline{DQS}$  signals must be monotonic between  $V_{IL}(DC)$ max and  $V_{IH}(DC)$ min. See Figure 18.
- 21. Input waveform timing tDH with differential data strobe enabled MR[bit10]=0, is referenced from the differential data strobe crosspoint to the input signal crossing at the  $V_{IH}(DC)$  level for a falling signal and from the differential data strobe crosspoint to the input signal crossing at the  $V_{IL}(DC)$  level for a rising signal applied to the device under test. DQS,  $\overline{DQS}$  signals must be monotonic between  $V_{IL}(DC)$ max and  $V_{IH}(DC)$ min. See Figure 18.

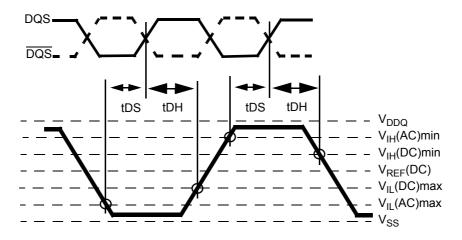


Figure 18 - Differential input waveform timing - tDS and tDH

- 22. Input waveform timing is referenced from the input signal crossing at the  $V_{IH}(AC)$  level for a rising signal and  $V_{IL}(AC)$  for a falling signal applied to the device under test. See Figure 19.
- 23. Input waveform timing is referenced from the input signal crossing at the  $V_{IL}(DC)$  level for a rising signal and  $V_{IH}(DC)$  for a falling signal applied to the device under test. See Figure 19.

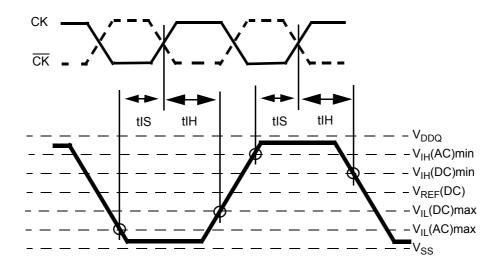


Figure 19 - Differential input waveform timing - tIS and tIH

- 24. tWTR is at lease two clocks (2 x tCK or 2 x nCK) independent of operation frequency.
- 25. Input waveform timing with single-ended data strobe enabled MR[bit10] = 1, is referenced from the input signal crossing at the  $V_{IH}(AC)$  level to the single-ended data strobe crossing  $V_{IH/L}(DC)$  at the start of its transition for a rising signal, and from the input signal crossing at the  $V_{IL}(AC)$  level to the single-ended data strobe crossing  $V_{IH/L}(DC)$  at the start of its transition for a falling signal applied to the device under test. The DQS signal must be monotonic between  $V_{IL}(DC)$ max and  $V_{IH}(DC)$ min.
- 26. Input waveform timing with single-ended data strobe enabled MR[bit10] = 1, is referenced from the input signal crossing at the  $V_{IH}(DC)$  level to the single-ended data strobe crossing  $V_{IH/L}(AC)$  at the end of its transition for a rising signal, and from the input signal crossing at the  $V_{IL}(DC)$  level to the single-ended data strobe crossing  $V_{IH/L}(AC)$  at the end of its transition for a falling signal applied to the device under test. The DQS signal must be monotonic between  $V_{II}(DC)$ max and  $V_{IH}(DC)$ min.
- 27. tCKEmin of 3 clocks means CKE must be registered on three consecutive positive clock edges. CKE must remain at the valid input level the entire time it takes to achieve the 3 clocks of registration. Thus, after any CKE transition, CKE may not transition from its valid level during the time period of tIS + 2 x tCK + tIH.
- 28. If tDS or tDH is violated, data corruption may occur and the data must be re-written with valid data before a valid READ can be executed.
- 29. These parameters are measured from a command/address signal (CKE,  $\overline{CS}$ ,  $\overline{RAS}$ ,  $\overline{CAS}$ ,  $\overline{WE}$ , ODT, BAO, AO, A1, etc.) transition edge to its respective clock signal (CK/ $\overline{CK}$ ) crossing. The spec values are not affected by the amount of clock jitter applied (i.e. tJIT(per), tJIT(cc), etc.), as the setup and hold are relative to the clock signal crossing that latches the command/address. That is, these parameters should be met whether clock jitter is present or not.
- **30.** These parameters are measured from a data strobe signal ((L/U/R)DQS/DQS) crossing to its respective clock signal (CK/CK) crossing. The spec values are not affected by the amount of clock jitter applied (i.e. tJIT(per), tJIT(cc), etc.), as these are relative to the clock signal crossing. That is, these parameters should be met whether clock jitter is present or not.
- 31. These parameters are measured from a data signal ((L/U)DM, (L/U)DQ0, (L/U)DQ1, etc.) transition edge to its respective data strobe signal ((L/U/R)DQS/DQS) crossing.
- 32. For these parameters, the DDR2 SDRAM device is characterized and verified to support tnPARAM = RU{tPARAM / tCK(avg)}, which is in clock cycles, assuming all input clock jitter specifications are satisfied.
  - For example, the device will support tnRP = RU{tRP / tCK(avg)}, which is in clock cycles, if all input clock jitter specifications are met. This means: For DDR2-667 5-5-5, of which tRP = 15ns, the device will support tnRP = RU{tRP / tCK(avg)} = 5, i.e. as long as the input clock jitter specifications are met, Precharge command at Tm and Active command at Tm+5 is valid even if (Tm+5 Tm) is less than 15ns due to input clock jitter.
- 33. tDAL [nCK] = WR [nCK] + tnRP [nCK] = WR + RU {tRP [ps] / tCK(avg) [ps] }, where WR is the value programmed in the mode register set.
- **34.** New units, 'tCK(avg)' and 'nCK', are introduced in DDR2-667 and DDR2-800. Unit 'tCK(avg)' represents the actual tCK(avg) of the input clock under operation. Unit 'nCK' represents one clock cycle of the input clock, counting the actual clock edges.

Note that in DDR2-400 and DDR2-533, 'tCK' is used for both concepts.

- ex) tXP = 2 [nCK] means; if Power Down exit is registered at Tm, an Active command may be registered at Tm+2, even if (Tm+2 Tm) is 2 x tCK(avg) + tERR(2per),min.
- 35. Input clock jitter spec parameter. These parameters and the ones in the table below are referred to as 'input clock jitter spec parameters' and these parameters apply to DDR2-667 and DDR2-800 only. The jitter specified is a random jitter meeting a Gaussian distribution.

Barrantar	Symbol	DDR2-667		DDR2-800			Natas
Parameter		Min	Max	Min	Max	units	Notes
Clock period jitter	tJIT(per)	-125	125	-100	100	ps	35
Clock period jitter during DLL locking period	tJIT(per,lck)	-100	100	-80	80	ps	35
Cycle to cycle clock period jitter	tJIT(cc)	-250	250	-200	200	ps	35
Cycle to cycle clock period jitter during DLL locking period	tJIT(cc,lck)	-200	200	-160	160	ps	35
Cumulative error across 2 cycles	tERR(2per)	-175	175	-150	150	ps	35
Cumulative error across 3 cycles	tERR(3per)	-225	225	-175	175	ps	35
Cumulative error across 4 cycles	tERR(4per)	-250	250	-200	200	ps	35
Cumulative error across 5 cycles	tERR(5per)	-250	250	-200	200	ps	35
Cumulative error across n cycles, n = 6 10, inclusive	tERR(6-10per)	-350	350	-300	300	ps	35
Cumulative error across n cycles, n = 11 50, inclusive	tERR(11-50per)	-450	450	-450	450	ps	35
Duty cycle jitter	tJIT(duty)	-125	125	-100	100	ps	35



#### **Definitions:**

- tCK(avg)

tCK(avg) is calculated as the average clock period across any consecutive 200 cycle window.

$$tCK(avg) = \left(\sum_{j=1}^{N} tCK_{j}\right)/N$$
where  $N = 200$ 

- tCH(avg) and tCL(avg)

tCH(avg) is defined as the average HIGH pulse width, as calculated across any consecutive 200 HIGH pulses.

$$tCH(avg) = \left(\sum_{j=1}^{N} tCH_{j}\right) / (N \times tCK(avg))$$
where  $N = 200$ 

tCL(avg) is defined as the average LOW pulse width, as calculated across any consecutive 200 LOW pulses.

$$tCL(avg) = \left(\sum_{j=1}^{N} tCL_{j}\right) / (N \times tCK(avg))$$
where  $N = 200$ 

- tJIT(duty)

tJIT(duty) is defined as the cumulative set of tCH jitter and tCL jitter. tCH jitter is the largest deviation of any single tCH from tCH(avg). tCL jitter is the largest deviation of any single tCL from tCL(avg).

tJIT(duty) = Min/max of {tJIT(CH), tJIT(CL)} where, tJIT(CH) = {tCHi- tCH(avg) where i=1 to 200}

tJIT(CL) = {tCLi- tCL(avg) where i=1 to 200}

- tJIT(per), tJIT(per,lck)

tJIT(per) is defined as the largest deviation of any single tCK from tCK(avg).

tJIT(per) = Min/max of {tCKi- tCK(avg) where i=1 to 200}

tJIT(per) defines the single period jitter when the DLL is already locked.

tJIT(per,lck) uses the same definition for single period jitter, during the DLL locking period only.

tJIT(per) and tJIT(per,lck) are not guaranteed through final production testing.

- tJIT(cc), tJIT(cc,lck)

tJIT(cc) is defined as the difference in clock period between two consecutive clock cycles: tJIT(cc) = Max of  $|tCK_{i+1} - tCKi|$ 

tJIT(cc) defines the cycle to cycle jitter when the DLL is already locked.

tJIT(cc,lck) uses the same definition for cycle to cycle jitter, during the DLL locking period only.

 $tJIT(cc) \ and \ tJIT(cc,lck) \ are \ not \ guaranteed \ through \ final \ production \ testing.$ 

- tERR(2per), tERR (3per), tERR (4per), tERR (5per), tERR (6-10per) and tERR (11-50per)
 tERR is defined as the cumulative error across multiple consecutive cycles from tCK(avg).

tERR(nper) = 
$$\begin{pmatrix} \sum_{j=1}^{n-1} tCK_j \\ j=1 \end{pmatrix} - n \times tCK(avg)$$
where 
$$\begin{pmatrix} n=2 & for & tERR(2per) \\ n=3 & for & tERR(3per) \\ n=4 & for & tERR(4per) \\ n=5 & for & tERR(5per) \\ 6 \le n \le 10 & for & tERR(6-10per) \\ 11 \le n \le 50 & for & tERR(11-50per) \end{pmatrix}$$

**36.** These parameters are specified per their average values, however it is understood that the following relationship between the average timing and the absolute instantaneous timing holds at all times. (Min and max of SPEC values are to be used for calculations in the table below.)

Parameter	Symbol	Min	Max	Units
Absolute clock Period	tCK(abs)	tCK(avg),min + tJIT(per),min	tCK(avg),max + tJIT(per),max	ps
Absolute clock HIGH pulse width	tCH(abs)	tCH(avg),min x tCK(avg),min + tJIT(duty),min	tCH(avg),max x tCK(avg),max + tJIT(duty),max	ps
Absolute clock LOW pulse width	tCL(abs)	tCL(avg),min x tCK(avg),min + tJIT(duty),min	tCL(avg),max x tCK(avg),max + tJIT(duty),max	ps

Example: For DDR2-667, tCH(abs),min = (0.48 x 3000 ps) - 125 ps = 1315 ps

37. tHP is the minimum of the absolute half period of the actual input clock. tHP is an input parameter but not an input specification parameter. It is used in conjunction with tQHS to derive the DRAM output timing tQH. The value to be used for tQH calculation is determined by the following equation; tHP = Min ( tCH(abs), tCL(abs)),

where.

tCH(abs) is the minimum of the actual instantaneous clock HIGH time;

tCL(abs) is the minimum of the actual instantaneous clock LOW time;

- 38. tQHS accounts for:
  - 1) The pulse duration distortion of on-chip clock circuits, which represents how well the actual tHP at the input is transferred to the output; and
- 2) The worst case push-out of DQS on one transition followed by the worst case pull-in of DQ on the next transition, both of which are independent of each other, due to data pin skew, output pattern effects, and p-channel to n-channel variation of the output drivers
- 39. tQH = tHP tQHS, where:

tHP is the minimum of the absolute half period of the actual input clock; and tQHS is the specification value under the max column.

{The less half-pulse width distortion present, the larger the tQH value is; and the larger the valid data eye will be.} Examples:

- 1) If the system provides tHP of 1315 ps into a DDR2-667 SDRAM, the DRAM provides tQH of 975 ps minimum.
- 2) If the system provides tHP of 1420 ps into a DDR2-667 SDRAM, the DRAM provides tQH of 1080 ps minimum.
- **40.** When the device is operated with input clock jitter, this parameter needs to be derated by the actual tERR(6-10per) of the input clock. (output deratings are relative to the SDRAM input clock.)

For example, if the measured jitter into a DDR2-667 SDRAM has tERR(6-10per),min = - 272 ps and tERR(6-10per),max = + 293 ps, then tDQSCK,min(derated) = tDQSCK,min - tERR(6-10per),max = - 400 ps - 293 ps = - 693 ps and tDQSCK,max(derated) = tDQSCK,max - tERR(6-10per),min = 400 ps + 272 ps = + 672 ps. Similarly, tLZ(DQ) for DDR2-667 derates to tLZ(DQ),min(derated) = - 900 ps - 293 ps = - 1193 ps and tLZ(DQ),max(derated) = 450 ps + 272 ps = + 722 ps.

**41.** When the device is operated with input clock jitter, this parameter needs to be derated by the actual tJIT(per) of the input clock. (output deratings are relative to the SDRAM input clock.)

For example, if the measured jitter into a DDR2-667 SDRAM has tJIT(per), min = -72 ps and tJIT(per), max = +93 ps, then tRPRE, min(derated) = tRPRE, min + tJIT(per), min = 0.9 x tCK(avg) - 72 ps = +2178 ps and tRPRE, max(derated) = tRPRE, max + tJIT(per), max = 1.1 x tCK(avg) + 93 ps = +2843 ps.

**42.** When the device is operated with input clock jitter, this parameter needs to be derated by the actual tJIT(duty) of the input clock. (output deratings are relative to the SDRAM input clock.)

For example, if the measured jitter into a DDR2-667 SDRAM has tJIT(duty), min = -72 ps and tJIT(duty), max = +93 ps, then tRPST, min + tJIT(duty),  $min = 0.4 \times tCK(avg) - 72$  ps = +928 ps and tRPST, max(derated) = tRPST, max + tJIT(duty),  $max = 0.6 \times tCK(avg) + 93$  ps = +1592 ps.

**43.** When the device is operated with input clock jitter, this parameter needs to be derated by { - tJIT(duty),max - tERR(6-10per),max } and { - tJIT(duty),min - tERR(6-10per),min } of the actual input clock. (output deratings are relative to the SDRAM input clock.)

For example, if the measured jitter into a DDR2-667 SDRAM has tERR(6-10per), min = -272 ps, tERR(6-10per), max = +293 ps, tJIT(duty), min = -106 ps and tJIT(duty), max = +94 ps, then tAOF, min(derated) = tAOF,  $min + {-tJIT(duty)}$ , max - tERR(6-10per),  $max } = -450$  ps  $+ {-50}$  ps  $+ {-50}$  ps and tAOF, max(derated) = tAOF,  $max + {-tJIT(duty)}$ , min - tERR(6-10per),  $min + {-100}$  ps  $+ {-100}$  ps +



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DDR2 1.55V SDRAM

**44.** For tAOFD of DDR2-400/533, the 1/2 clock of tCK in the 2.5 x tCK assumes a tCH, input clock HIGH pulse width of 0.5 relative to tCK. tAOF,min and tAOF,max should each be derated by the same amount as the actual amount of tCH offset present at the DRAM input with respect to 0.5.

For example, if an input clock has a worst case tCH of 0.45, the tAOF,min should be derated by subtracting 0.05 x tCK from it, whereas if an input clock has a worst case tCH of 0.55, the tAOF,max should be derated by adding 0.05 x tCK to it. Therefore, we have;

tAOF,min(derated) = tAC,min - [0.5 - Min(0.5, tCH,min)] x tCK

tAOF,max(derated) = tAC,max + 0.6 + [Max(0.5, tCH,max) - 0.5] x tCK

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tAOF,min(derated) = Min(tAC,min, tAC,min - [0.5 - tCH,min] x tCK)

tAOF,max(derated) = 0.6 + Max(tAC,max, tAC,max + [tCH,max - 0.5] x tCK)

where tCH,min and tCH,max are the minimum and maximum of tCH actually measured at the DRAM input balls.

**45.** For tAOFD of DDR2-667/800, the 1/2 clock of nCK in the 2.5 x nCK assumes a tCH(avg), average input clock HIGH pulse width of 0.5 relative to tCK(avg). tAOF,min and tAOF,max should each be derated by the same amount as the actual amount of tCH(avg) offset present at the DRAM input with respect to 0.5.

For example, if an input clock has a worst case tCH(avg) of 0.48, the tAOF,min should be derated by subtracting 0.02 x tCK(avg) from it, whereas if an input clock has a worst case tCH(avg) of 0.52, the tAOF,max should be derated by adding 0.02 x tCK(avg) to it. Therefore, we have;

tAOF,min(derated) = tAC,min - [0.5 - Min(0.5, tCH(avg),min)] x tCK(avg)

tAOF,max(derated) = tAC,max + 0.6 + [Max(0.5, tCH(avg),max) - 0.5] x tCK(avg)

tAOF,min(derated) = Min(tAC,min, tAC,min - [0.5 - tCH(avg),min] x tCK(avg))

tAOF,max(derated) = 0.6 + Max(tAC,max, tAC,max + [tCH(avg),max - 0.5] x tCK(avg))

where tCH(avg),min and tCH(avg),max are the minimum and maximum of tCH(avg) actually measured at the DRAM input balls.

Note that these deratings are in addition to the tAOF derating per input clock jitter, i.e. tJIT(duty) and tERR(6-10per). However tAC values used in the equations shown above are from the timing parameter table and are not derated. Thus the final derated values for tAOF are;

tAOF,min(derated\_final) = tAOF,min(derated) + { - tJIT(duty),max - tERR(6-10per),max }

tAOF,max(derated\_final) = tAOF,max(derated) + { - tJIT(duty),min - tERR(6-10per),min }

