

# AH312

2 Watt, High Linearity InGaP HBT Amplifier



## Product Features

- 400 – 2300 MHz
- +33 dBm P1dB
- +51 dBm Output IP3
- 18 dB Gain @ 900 MHz
- +5V Single Positive Supply
- MTTF > 100 Years
- Lead-free/RoHS-compliant SOIC-8 SMT Pkg.

## Applications

- Final stage amplifiers for Repeaters
- Mobile Infrastructure

## Specifications <sup>(1)</sup>

Parameter	Units	Min	Typ	Max
Operational Bandwidth	MHz	400		2300
Test Frequency	MHz		2140	
Gain	dB	9	10	
Input R.L.	dB		20	
Output R.L.	dB		6.8	
Output P1dB	dBm	+32	+33.2	
Output IP3 <sup>(2)</sup>	dBm	+47	+48	
wCDMA Channel Power @ -45 dBc ACLR, 2140 MHz	dBm		+25.3	
Noise Figure	dB		7.7	
Operating Current Range, Icc <sup>(3)</sup>	mA	700	800	900
Device Voltage, Vcc	V		+5	

1. Test conditions unless otherwise noted: 25°C, +5V V<sub>supply</sub>, 2140 MHz, in tuned application circuit.

2. 3OIP measured with two tones at an output power of +17 dBm/tone separated by 1 MHz. The suppression on the largest IM3 product is used to calculate the 3OIP using a 2:1 rule.

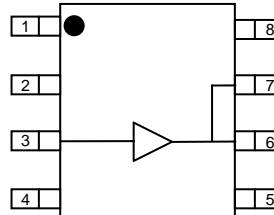
3. This corresponds to the quiescent current or operating current under small-signal conditions into pins 6, 7, and 8. It is expected that the current can increase by an additional 200 mA at P1dB. Pin 1 is used as a reference voltage for the internal biasing circuitry. It is expected that Pin 1 will pull 22mA of current when used with a series bias resistor of R1=15Ω. (ie. total device current typically will be 822 mA.)

## Product Description

The AH312 is a high dynamic range driver amplifier in a low-cost surface mount package. The InGaP/GaAs HBT is able to achieve high performance for various narrowband-tuned application circuits with up to +49 dBm OIP3 and +33 dBm of compressed 1dB power. It is housed in a lead-free/RoHS-compliant SOIC-8 package. All devices are 100% RF and DC tested.

The AH312 is targeted for use as a driver amplifier in wireless infrastructure where high linearity and medium power is required. An internal active bias allows the AH312 to maintain high linearity over temperature and operate directly off a single +5V supply. This combination makes the device an excellent candidate for transceiver line cards in current and next generation multi-carrier 3G base stations.

## Functional Diagram



Function	Pin No.
Vref	1
Input	3
Output	6, 7
Vbias	8
GND	Backside Paddle
N/C or GND	2, 4, 5

## Typical Performance <sup>(4)</sup>

Parameter	Units	Typical		
Frequency	MHz	900	1960	2140
S21 – Gain	dB	18	11	10
S11 – Input R.L.	dB	-18	-19	-20
S22 – Output R.L.	dB	-11	-6.8	-6.8
Output P1dB	dBm	+33	+33.4	+33.2
Output IP3	dBm	+49	+51	+48
IS-95A Channel Power @ -45 dBc ACPR	dBm	+27	+27.5	
wCDMA Channel Power @ -45 dBc ACLR	dBm			+25.3
Noise Figure	dB	8.0	7.3	7.7
Device Bias <sup>(3)</sup>				+5 V @ 800 mA

4. Typical parameters reflect performance in a tuned application circuit at +25°C.

## Not Recommended for New Designs

Recommended Replacement  
Part: TQP7M9104

## Absolute Maximum Rating

Parameter	Rating
Thermal Resistance, R <sub>th</sub>	17.5°C/W
Storage Temperature	-65 to +150 °C
RF Input Power (continuous)	+28 dBm
Device Voltage	+8 V
Device Current	1400 mA
Device Power	8 W
Junction Temperature	+200 °C

Operation of this device above any of these parameters may cause permanent damage.

Specifications and information are subject to change without notice.

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## Ordering Information

Part No.	Description
AH312-S8G	2 Watt, High Linearity InGaP HBT Amplifier (lead-free/RoHS-compliant SOIC-8 Pkg)

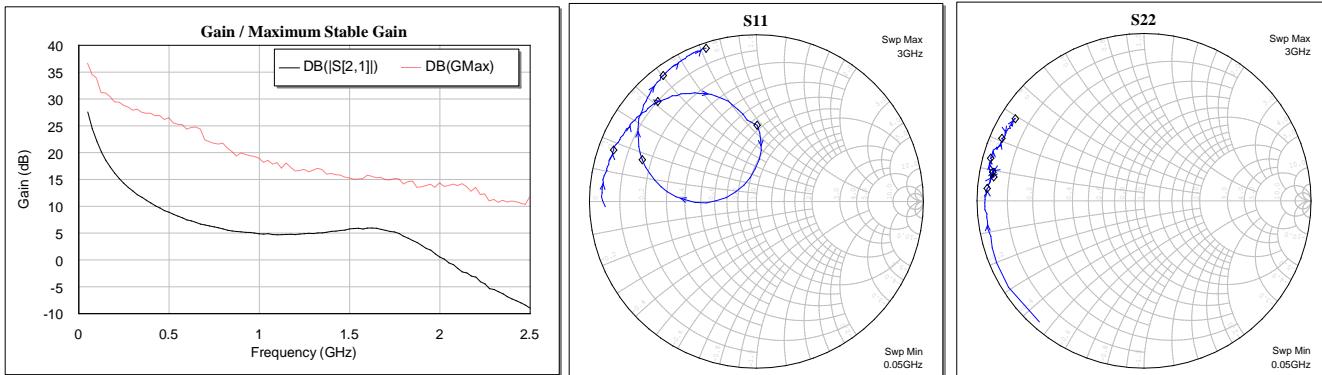
Standard tape / reel size = 500 pieces on a 7" reel

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### Typical Device Data

S-Parameters ( $V_{CC} = +5$  V,  $I_{CC} = 800$  mA,  $T = 25$  °C, calibrated to device leads)



Notes:

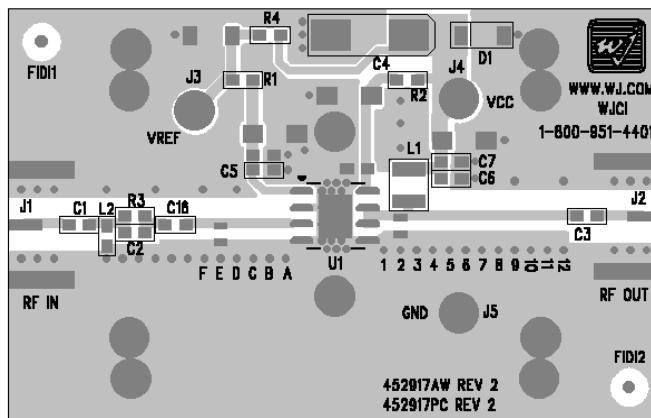
The gain for the unmatched device in 50 ohm system is shown as the trace in black color. For a tuned circuit for a particular frequency, it is expected that actual gain will be higher, up to the maximum stable gain. The maximum stable gain is shown in the dashed red line. The impedance plots are shown from 50 – 3000 MHz, with markers placed at 0.5 – 3.0 GHz in 0.5 GHz increments.

S-Parameters ( $V_{CC} = +5$  V,  $I_{CC} = 800$  mA,  $T = 25$  °C, unmatched 50 ohm system, calibrated to device leads)

Freq (MHz)	$S_{11}$ (dB)	$S_{11}$ (ang)	$S_{21}$ (dB)	$S_{21}$ (ang)	$S_{12}$ (dB)	$S_{12}$ (ang)	$S_{22}$ (dB)	$S_{22}$ (ang)
50	-0.86	-178.06	27.55	113.72	-45.75	30.91	-0.38	-130.98
100	-0.64	178.18	22.16	98.81	-45.46	12.80	-0.38	-157.30
200	-0.68	172.85	16.13	89.06	-42.65	6.09	-0.48	-172.51
400	-0.76	164.33	10.61	77.31	-43.96	4.69	-0.48	177.51
600	-0.93	155.56	7.46	67.94	-41.17	6.70	-0.61	173.63
800	-1.15	146.04	5.78	57.62	-41.65	-5.78	-0.66	170.49
1000	-1.50	134.58	4.87	46.90	-40.36	-7.84	-0.71	169.31
1200	-2.39	121.66	4.74	32.96	-40.22	-16.51	-0.80	168.22
1400	-4.47	104.01	5.33	14.01	-38.97	-48.82	-0.76	167.91
1600	-11.96	86.06	5.96	-17.55	-38.96	-86.32	-0.60	170.63
1800	-8.66	-179.11	4.41	-56.78	-39.35	-144.53	-0.52	167.41
2000	-2.76	159.91	0.53	-89.86	-43.55	145.94	-0.41	164.50
2200	-1.21	142.90	-3.21	-107.99	-41.56	104.25	-0.54	160.11
2400	-0.68	130.93	-7.27	-123.14	-42.46	73.64	-0.68	157.84
2600	-0.43	121.91	-10.41	-134.93	-39.71	64.28	-0.73	154.66
2800	-0.32	114.61	-13.28	-143.22	-40.99	58.20	-0.73	151.14
3000	-0.29	108.16	-15.94	-149.93	-39.65	48.40	-0.79	147.52

Device S-parameters are available for download off of the website at: [www.TriQuint.com](http://www.TriQuint.com)

### Application Circuit PC Board Layout



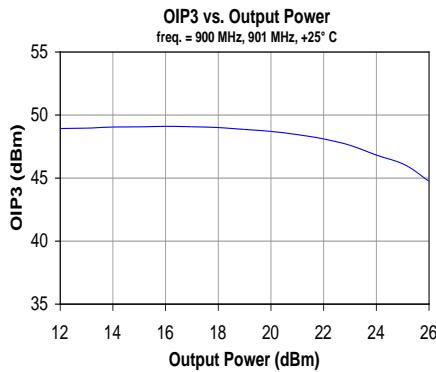
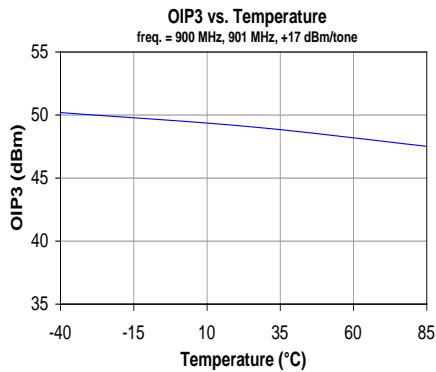
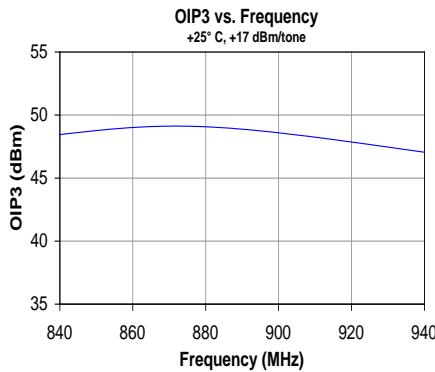
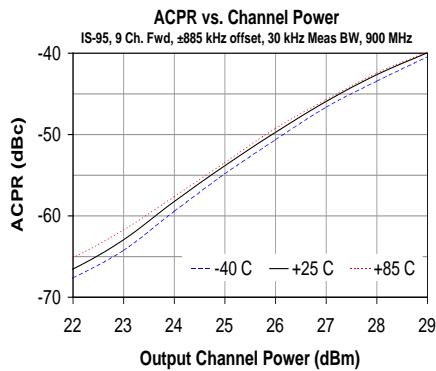
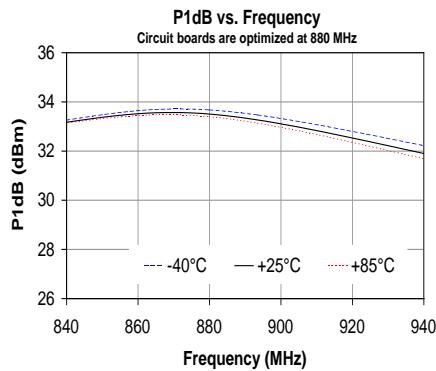
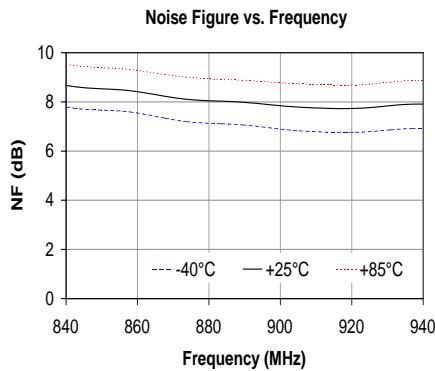
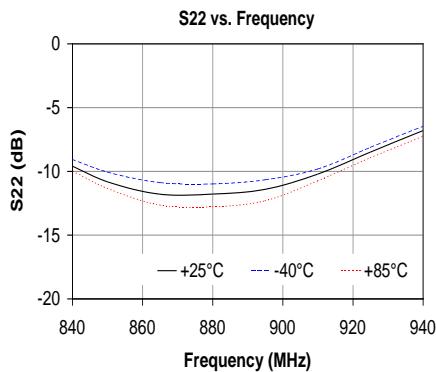
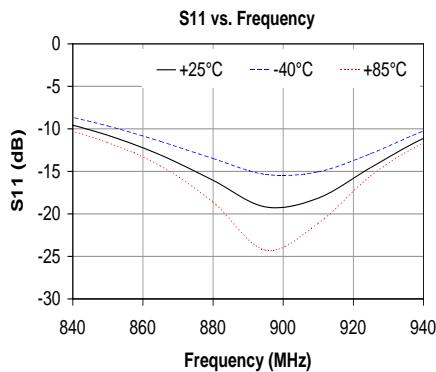
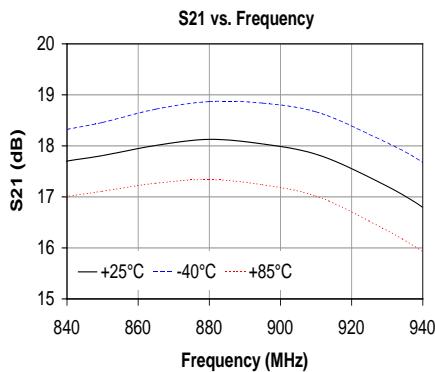
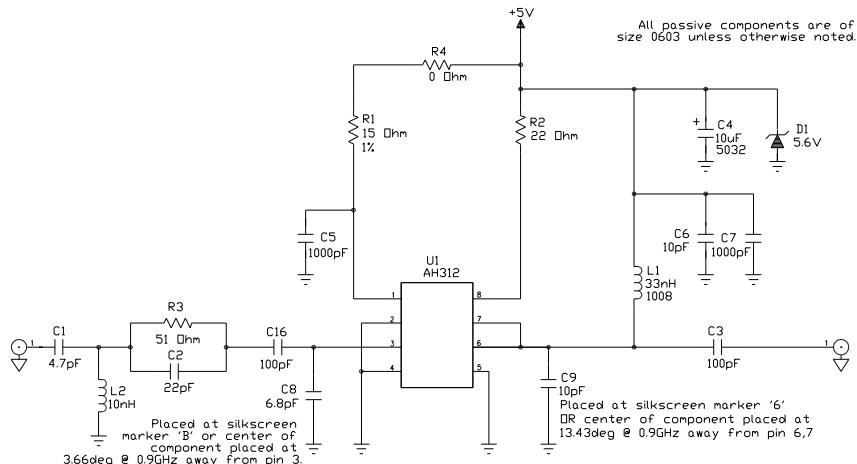
Circuit Board Material: .014" Getek, single layer, 1 oz copper, Microstrip line details: width = .026", spacing = .026". The silk screen markers 'A', 'B', 'C', etc. and '1', '2', '3', etc. are used as placemarkers for the input and output tuning shunt capacitors – C8 and C9. The markers and vias are spaced in .050" increments.

### 900 MHz Application Circuit (AH312-S8PCB900)

#### Typical RF Performance at 25 °C

Frequency	900 MHz
S21 – Gain	18 dB
S11 – Input Return Loss	-18 dB
S22 – Output Return Loss	-11 dB
Output P1dB	+33 dBm
Output IP3 (+17 dBm / tone, 1 MHz spacing)	+49 dBm
Channel Power (@-45 dBc ACPR, IS-95 9 channels fwd)	+27 dBm
Noise Figure	8.0 dB
Device / Supply Voltage	+5 V
Quiescent Current (I)	800 mA

1. This corresponds to the quiescent current or operating current under small-signal conditions into pins 6, 7, and 8.



# AH312

2 Watt, High Linearity InGaP HBT Amplifier

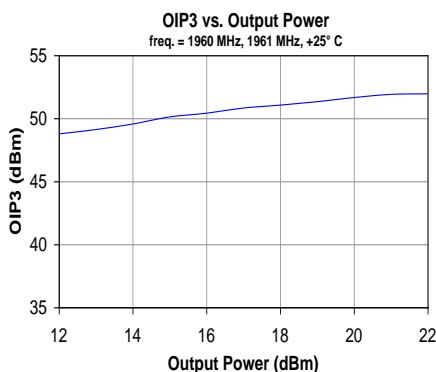
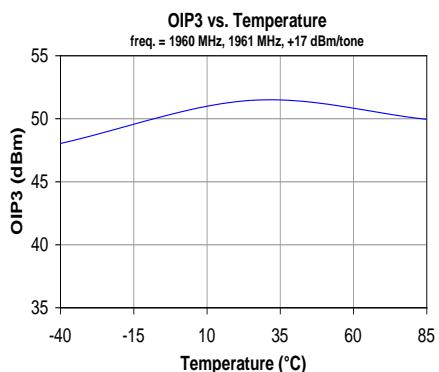
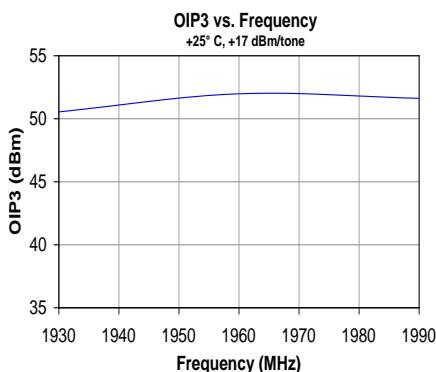
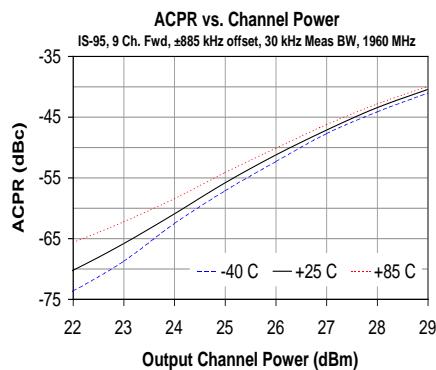
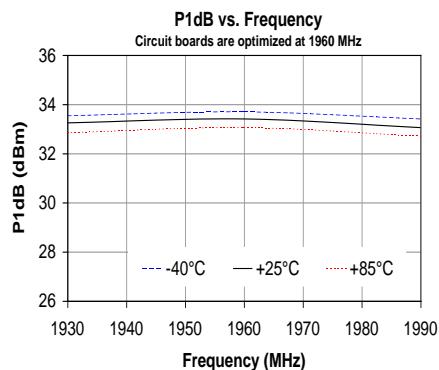
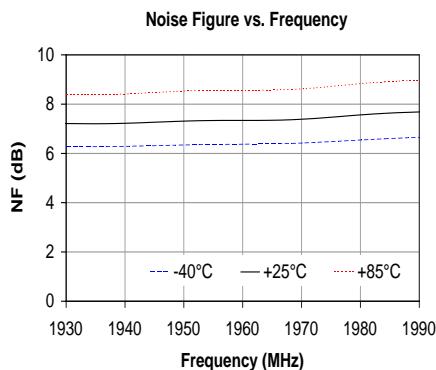
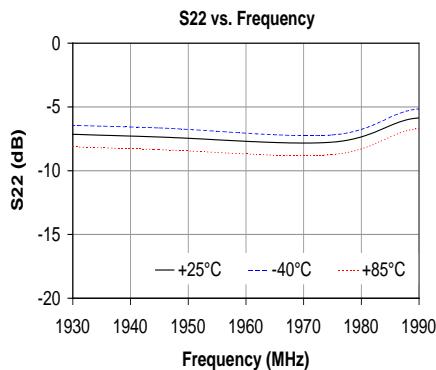
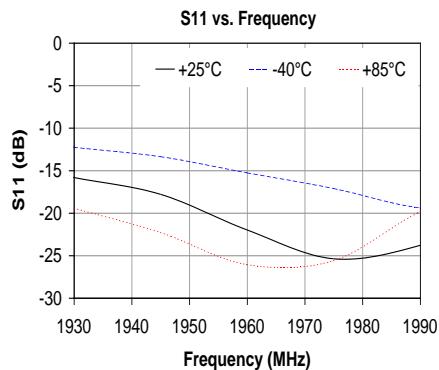
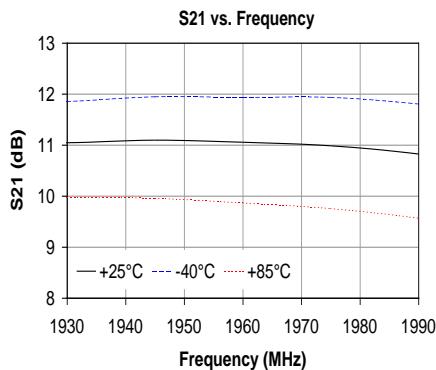
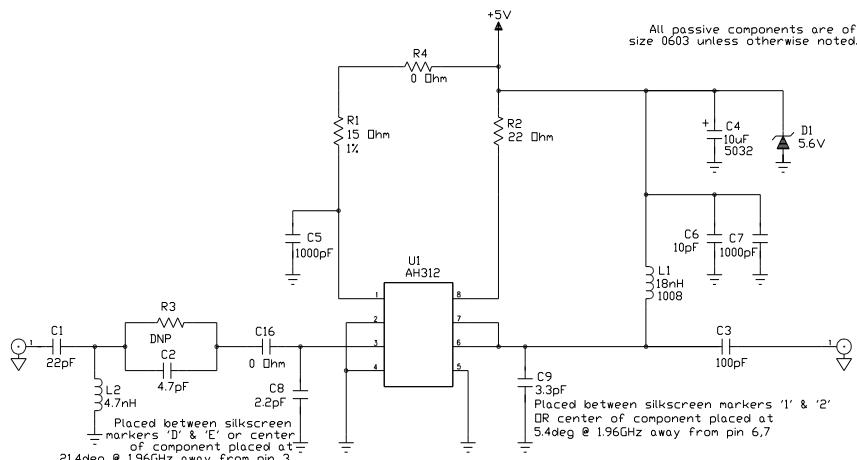
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## 1960 MHz Application Circuit (AH312-S8PCB1960)

### Typical RF Performance at 25 °C

Frequency	1960 MHz
S21 – Gain	11 dB
S11 – Input Return Loss	-20 dB
S22 – Output Return Loss	-6.8 dB
Output P1dB	+33.4 dBm
Output IP3 (+17 dBm / tone, 1 MHz spacing)	+51 dBm
Channel Power (@-45 dBc ACPR, IS-95 9 channels fwd)	+27.5 dBm
Noise Figure	7.3 dB
Device / Supply Voltage	+5 V
Quiescent Current (I <sub>Q</sub> )	800 mA

1. This corresponds to the quiescent current or operating current under small-signal conditions into pins 6, 7, and 8.



Specifications and information are subject to change without notice.

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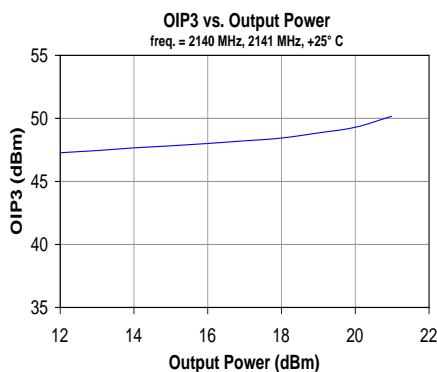
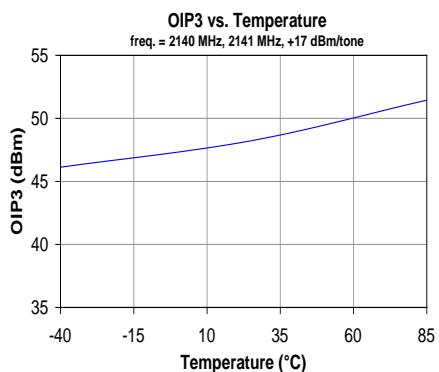
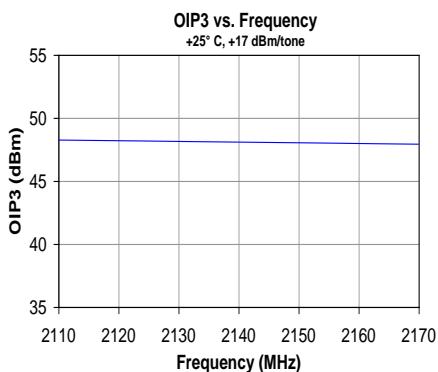
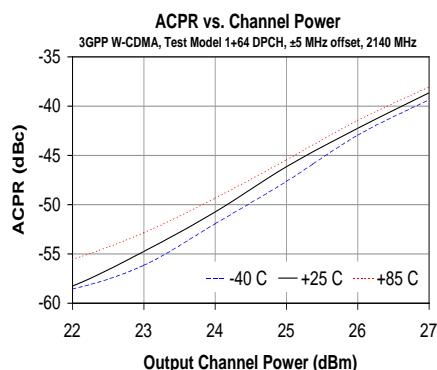
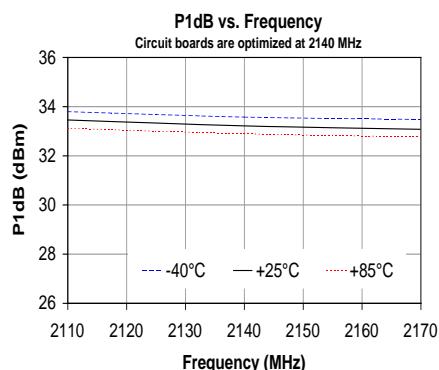
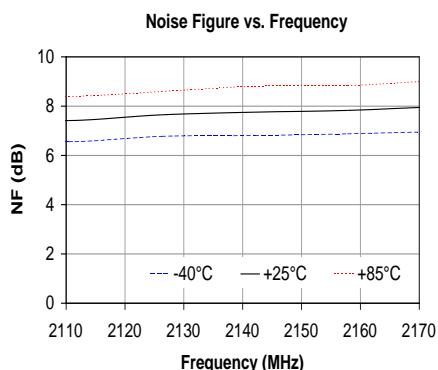
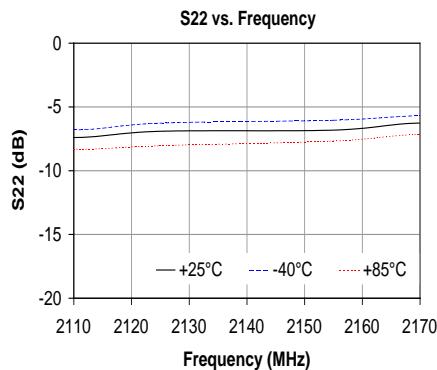
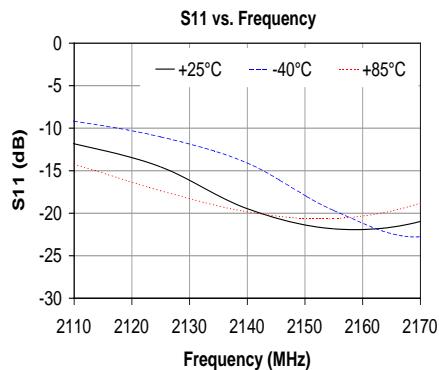
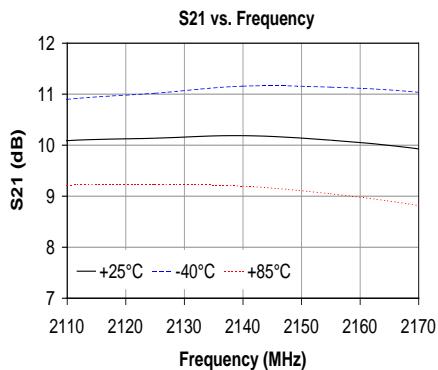
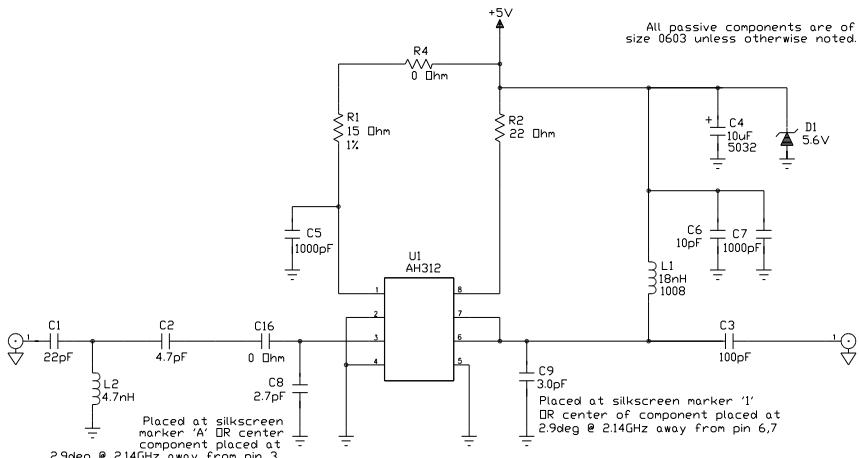
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### 2140 MHz Application Circuit (AH312-S8PCB2140)

#### Typical RF Performance at 25 °C

Frequency	2140 MHz
S21 – Gain	10 dB
S11 – Input Return Loss	-20 dB
S22 – Output Return Loss	-6.8 dB
Output P1dB	+33.2 dBm
Output IP3 (+17 dBm / tone, 1 MHz spacing)	+48 dBm
wCDMA Channel Power (@-45 dBc ACLR, 3GPP, TM 1+64 DPCH)	+25.3 dBm
Noise Figure	7.7 dB
Device / Supply Voltage	+5 V
Quiescent Current <sup>(1)</sup>	800 mA

1. This corresponds to the quiescent current or operating current under small-signal conditions into pins 6, 7, and 8.

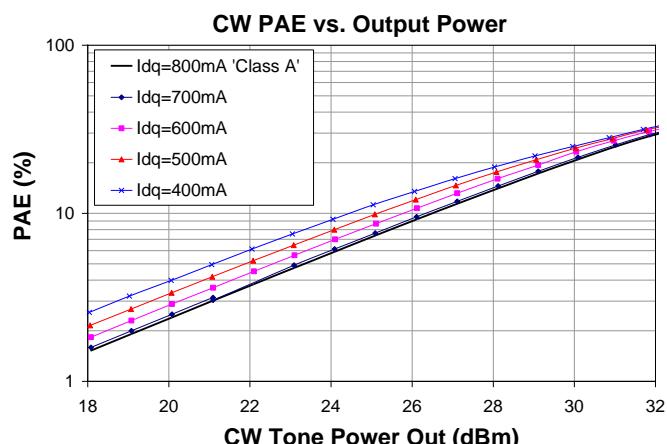
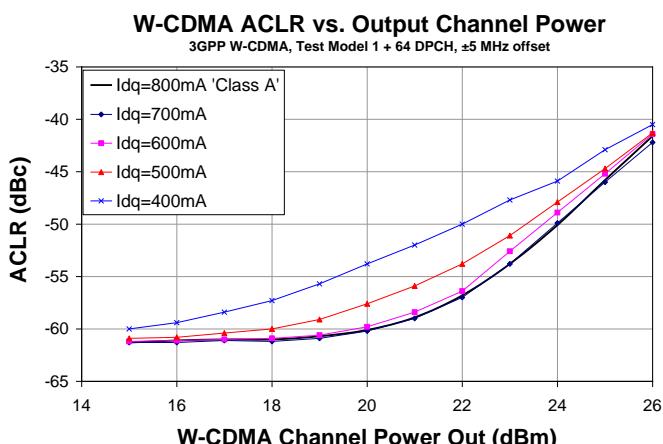
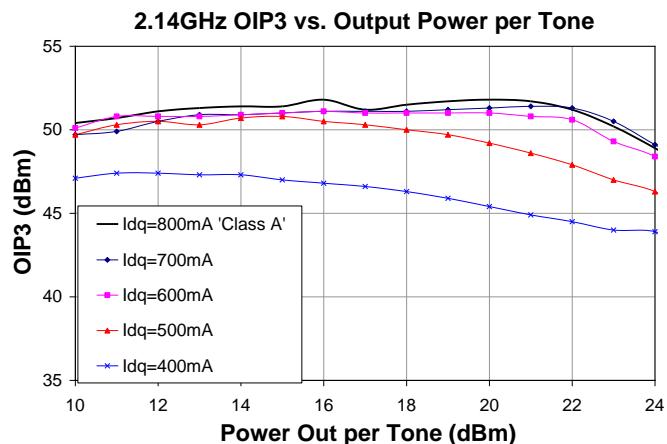
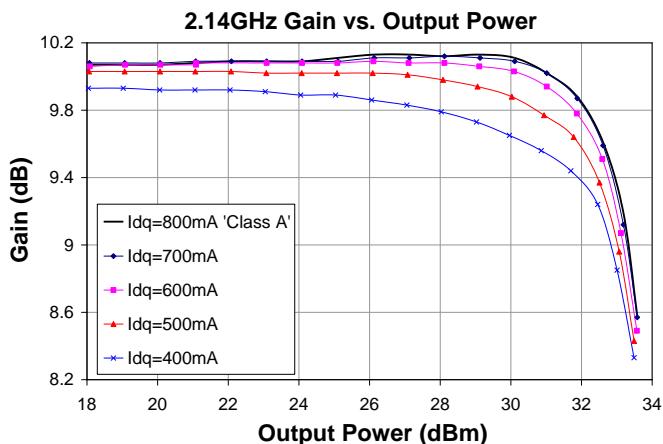
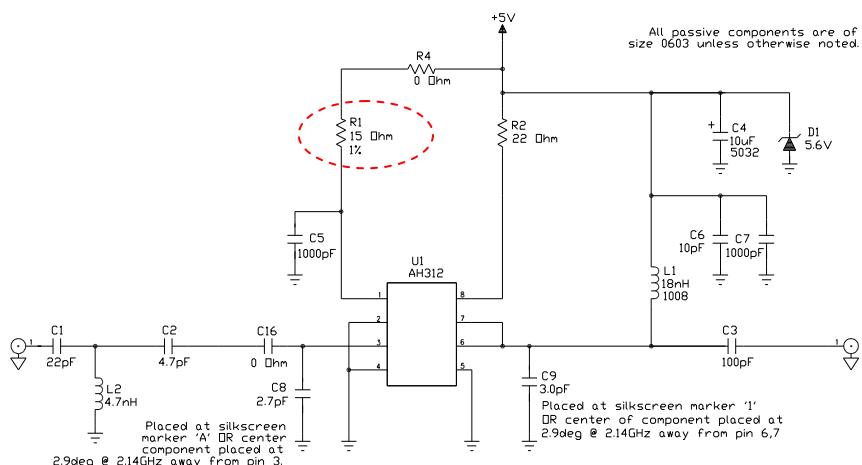


## Application Note: Reduced Bias Configurations

The AH312 can be configured to be operated with lower bias current by varying the bias-adjust resistor – R1. The recommended circuit configurations shown previously in this datasheet have the device operating in Class A operation. Lowering the current has little effect on the gain, OIP3, and P1dB performance of the device, but will slightly lower the ACLR/ACPR performance of the device as shown below. An example of the measured data below represents the AH312 measured and configured for 2.14 GHz applications. It is expected that variation of the bias current for other frequency applications will produce similar performance results.

### AH312-S8PCB2140 Performance Data

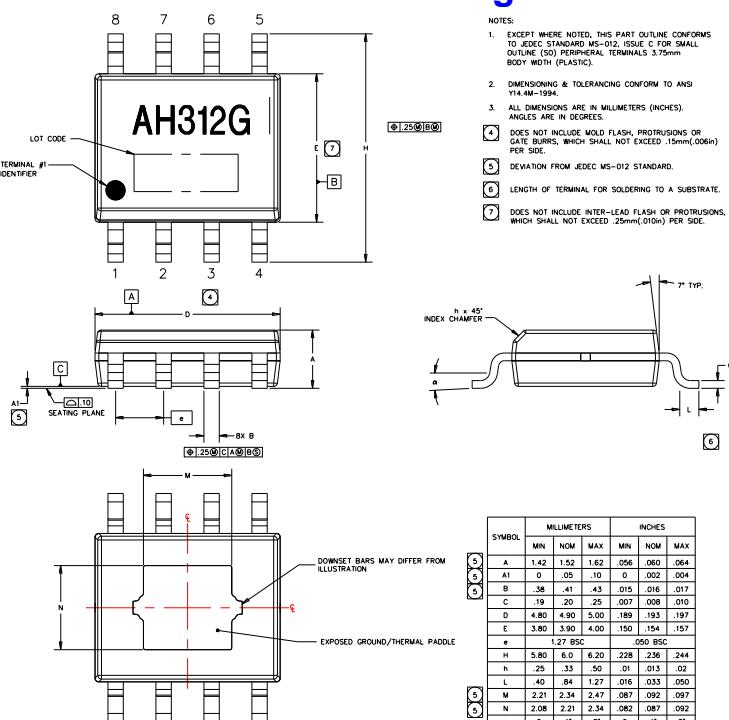
R1 (ohms)	Icq (mA)	Pdiss (W)	P1dB (dBm)	OIP3 (dBm)
15	800	4.0	+33.3	+51.4
22	700	3.5	+33.3	+50.9
43	600	3.0	+33.1	+50.9
62	500	2.5	+33.0	+50.7
110	400	2.0	+32.9	+47.3



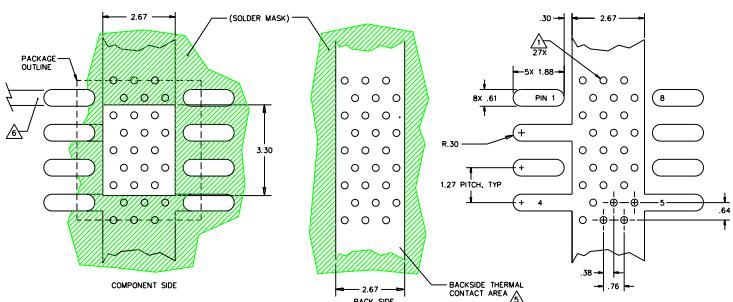
### AH312-S8G (Lead-Free Package) Mechanical Information

This package is lead-free/RoHS-compliant. The plating material on the leads is NiPdAu. It is compatible with both lead-free (maximum 260°C reflow temperature) and lead (maximum 245 °C reflow temperature) soldering processes.

#### Outline Drawing



#### Mounting Configuration / Land Pattern



#### Product Marking

The component will be marked with an "AH312G" designator with an alphanumeric lot code on the top surface of the package.

#### ESD / MSL Information



**Caution!** ESD sensitive device.

ESD Rating: Class 1B  
 Value: Passes between 500 and 1000V  
 Test: Human Body Model (HBM)  
 Standard: JEDEC Standard JESD22-A114

MSL Rating: Level 2 at +260 °C convection reflows  
 Standard: JEDEC Standard J-STD-020

#### Mounting Config. Notes

1. A heatsink underneath the area of the PCB for the mounted device is strictly required for proper thermal operation. Damage to the device can occur without the use of one.
2. Ground / thermal vias are critical for the proper performance of this device. Vias should use a .35mm (#80/.0135") diameter drill and have a final plated thru diameter of .25 mm (.010").
3. Add as much copper as possible to inner and outer layers near the part to ensure optimal thermal performance.
4. Mounting screws can be added near the part to fasten the board to a heatsink. Ensure that the ground / thermal via region contacts the heatsink.
5. Do not put solder mask on the backside of the PC board in the region where the board contacts the heatsink.
6. RF trace width depends upon the PC board material and construction.
7. Use 1 oz. Copper minimum.
8. All dimensions are in millimeters (inches). Angles are in degrees.