

## FEATURES

- **Ultralow Quiescent Current: 3.5 $\mu$ A Typ**
- **Open-Drain Outputs Typically Sink Greater Than 20mA**
- **Wide Supply Range: (LTC1841)**  
Single: 2V to 11V  
Dual:  $\pm 1$ V to  $\pm 5.5$ V
- **Input Voltage Range Includes the Negative Supply**
- **Reference Output Drives 0.01 $\mu$ F Capacitor**
- **Adjustable Hysteresis**
- **12 $\mu$ s Propagation Delay with 10mV Overdrive**
- **No Current Spike When Switching**

## APPLICATIONS

- Battery-Powered System Monitoring
- Threshold Detectors
- Window Comparators
- Oscillator Circuits

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## DESCRIPTION

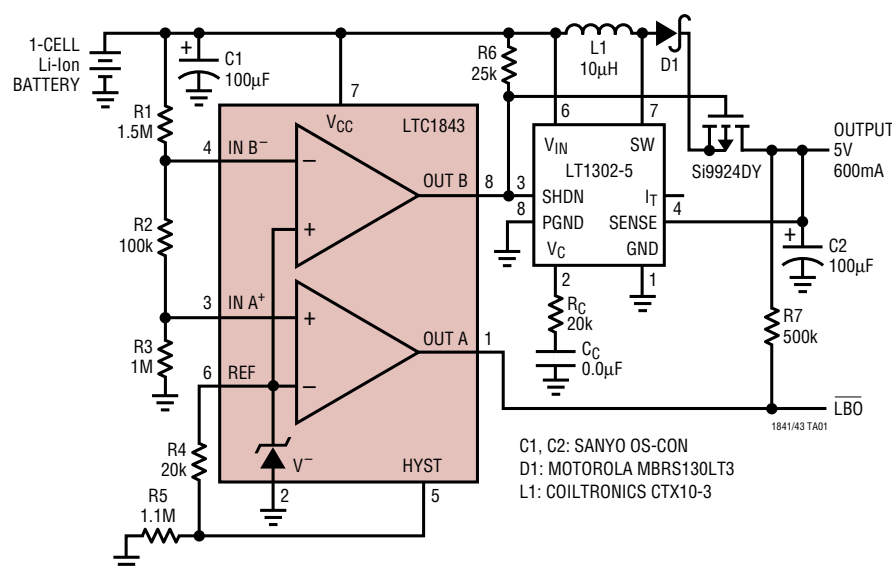
The LTC®1841/LTC1842/LTC1843 are ultralow power dual comparators with built-in reference (LTC1842/LTC1843). The comparators feature less than 5.7µA supply current over temperature, a 1.182V ±1% reference, programmable hysteresis and open-drain outputs that sink current. The reference output can drive a bypass capacitor of up to 0.01µF without oscillation.

The LTC1841 operates from a single 2V to 11V supply or a dual  $\pm 1\text{V}$  to  $\pm 5.5\text{V}$  supply. The LTC1842/LTC1843 operate from a single 2.5V to 11V supply or a dual  $\pm 1.25\text{V}$  to  $\pm 5.5\text{V}$  supply. The LTC1842/LTC1843 hysteresis is easily programmed by using two resistors and the HYST pin. The comparators' input operates from the negative supply to within 1.3V of the positive supply. The comparators' output stage can typically sink greater than 20mA. By eliminating the cross-conduction current that normally happens when the comparators change logic states, power supply glitches are eliminated.

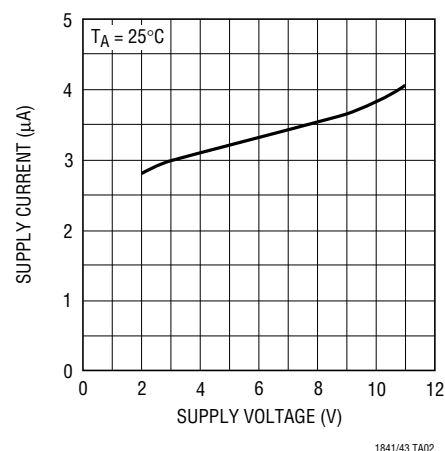
The LTC1841/LTC1842/LTC1843 are available in SO-8 packages.

## TYPICAL APPLICATION

## Single Li-Ion Cell to 5V Supply



### LTC1843 Supply Current vs Supply Voltage



# LTC1841/LTC1842/LTC1843

## ABSOLUTE MAXIMUM RATINGS

(Note 1)

Voltage

$V^+$ to $V^-$	12V to $-0.3V$
$IN^+$ , $IN^-$ , HYST	$(V^+ + 0.3V)$ to $(V^- - 0.3V)$
REF	$(V^+ + 0.3V)$ to $(V^- - 0.3V)$
OUT	12V to $(V^- - 0.3V)$

Current

$IN^+$ , $IN^-$ , HYST	20mA
REF	20mA
OUT	50mA

OUT Short-Circuit Duration ( $V^+ \leq 5.5V$ ) ..... Continuous  
Power Dissipation ..... 500mW

Operating Temperature Range

LTC1841C/LTC1842C/LTC1843C	$0^\circ C$ to $70^\circ C$
LTC1841I/LTC1842I/LTC1843I	$-40^\circ C$ to $85^\circ C$

Storage Temperature Range .....  $-65^\circ C$  to  $150^\circ C$

Lead Temperature (Soldering, 10 sec) .....  $300^\circ C$

## PACKAGE/ORDER INFORMATION

<p>TOP VIEW</p> <p>S8 PACKAGE 8-LEAD PLASTIC SO <math>T_{JMAX} = 150^\circ C</math>, <math>\theta_{JA} = 125^\circ C/W</math></p>		<p>TOP VIEW</p> <p>S8 PACKAGE 8-LEAD PLASTIC SO <math>T_{JMAX} = 150^\circ C</math>, <math>\theta_{JA} = 125^\circ C/W</math></p>		<p>TOP VIEW</p> <p>S8 PACKAGE 8-LEAD PLASTIC SO <math>T_{JMAX} = 150^\circ C</math>, <math>\theta_{JA} = 125^\circ C/W</math></p>	
ORDER PART NUMBER	S8 PART MARKING	ORDER PART NUMBER	S8 PART MARKING	ORDER PART NUMBER	S8 PART MARKING
LTC1841CS8 LTC1841IS8	1841 1841I	LTC1842CS8 LTC1842IS8	1842 1842I	LTC1843CS8 LTC1843IS8	1843 1843I

Consult factory for Military grade parts.

## ELECTRICAL CHARACTERISTICS $V^+ = 5V$ , $V^- = 0V$ , $T_A = 25^\circ C$ unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
<b>Power Supply</b>						
$V^+$	Supply Voltage Range	LTC1841	2.0		11	V
		LTC1842/LTC1843	2.5		11	V
$I_{CC}$	Supply Current	HYST = REF (Note 2)		3.5	5.7	$\mu A$
<b>Comparator</b>						
$V_{OS}$	Comparator Input Offset Voltage	(Note 3)		$\pm 3$	$\pm 10$	mV
$I_{IN}$	Input Leakage Current ( $IN^+$ , $IN^-$ )	$IN^+ = IN^- = 2.5V$ (LTC1841), $IN^+ = IN^- = V_{REF}$ (LTC1842/LTC1843)		$\pm 0.01$	$\pm 1.0$	nA
$V_{IN}$	Comparator Input Voltage Range		$V^-$		$V^+ - 1.3V$	V
PSRR	Power Supply Rejection Ratio	$V^+ = 2V$ to $11V$ (LTC1841), $V^+ = 2.5V$ to $11V$ (LTC1842/LTC1843)		0.1	1	mV/V
CMRR	Common Mode Rejection Ratio	$V_{CM} = V^-$ to $(V^+ - 1.3V)$ LTC1841		0.1	1	mV/V
$V_{HYST}$	Hysteresis Input Voltage Range	LTC1842/LTC1843	$V_{REF} - 50mV$		$V_{REF}$	V

**ELECTRICAL CHARACTERISTICS**  $V^+ = 5V$ ,  $V^- = 0V$ ,  $T_A = 25^\circ C$  unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
$t_{PD}$	Propagation Delay	$C_{OUT} = 10pF$ , $R_{PULL-UP} = 100k$ Overdrive = 10mV Overdrive = 100mV		12 4		$\mu s$ $\mu s$
$I_{LEAK}$	Output Leakage Current	$V_{OUT} = 12V$ (Note 2)	●	1	100	nA
$V_{OL}$	Output Low Voltage	$I_{OUT} = 1.8mA$	●		$V^- + 0.4V$	V

**Reference (LTC1842/LTC1843)**

$V_{REF}$	Reference Voltage	No Load	25°C		1.174	1.182	1.190	V
			0°C to 70°C	●	1.170		1.194	V
			-40°C to 85°C	●	1.164		1.200	V
$\Delta V_{REF}$	Load Regulation	$I_{SOURCE} = 1mA$	●		1		3	mV
		$I_{SINK} = 10\mu A$	●		1.75		4	mV
$e_n$	Voltage Noise	100Hz to 100kHz					6	mV
					100			$\mu V_{RMS}$

 $V^+ = 3V$ ,  $V^- = 0V$ ,  $T_A = 25^\circ C$  unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
Power Supply							
V <sup>+</sup>	Supply Voltage Range	LTC1841	●	2.0		11	V
		LTC1842/LTC1843	●	2.5		11	V
I <sub>CC</sub>	Supply Current	HYST = REF (Note 2)	●		3.5	5.5	μA

**Comparator**

$V_{OS}$	Comparator Input Offset Voltage	(Note 3)	●		$\pm 3$	$\pm 10$	mV
$I_{IN}$	Input Leakage Current ( $IN^+$ , $IN^-$ )	$IN^+ = IN^- = 1.5V$ (LTC1841), $IN^+ = IN^- = V_{REF}$ (LTC1842/LTC1843)	●		$\pm 0.01$	$\pm 1$	nA
$V_{IN}$	Comparator Input Voltage Range		●	$V^-$		$V^+ - 1.3V$	V
PSRR	Power Supply Rejection Ratio	$V^+ = 2V$ to 11V (LTC 1841), $V^+ = 2.5V$ to 11V (LTC1842/LTC1843)			0.1	1	mV/V
CMRR	Common Mode Rejection Ratio	$V_{CM} = V^-$ to $(V^+ - 1.3V)$ LTC1841	●		0.1	1	mV/V
$V_{HYST}$	Hysteresis Input Voltage Range	LTC1842/LTC1843	●	$V_{REF} - 50mV$		$V_{REF}$	V
$t_{PD}$	Propagation Delay	$C_{OUT} = 10pF$ , $R_{PULL-UP} = 100k$ Overdrive = 10mV Overdrive = 100mV			14 5		$\mu s$ $\mu s$
$I_{LEAK}$	Output Leakage Current	$V_{OUT} = 12V$ (Note 2)	●		1	100	nA
$V_{OL}$	Output Low Voltage	$I_O = 0.8mA$	●			$V^- + 0.4V$	V

**Reference (LTC1842/LTC1843)**

$V_{REF}$	Reference Voltage	No Load	25°C		1.174	1.182	1.190	V
			0°C to 70°C	●	1.170		1.194	V
			-40°C to 85°C	●	1.164		1.200	V
$\Delta V_{REF}$	Load Regulation	$I_{SOURCE} = 1mA$	●		2		6	mV
		$I_{SINK} = 10\mu A$	●		1.75		4	mV
$e_n$	Voltage Noise	100Hz to 100kHz					6	mV
					100			$\mu V_{RMS}$

The ● denotes specifications which apply over the full operating temperature range.

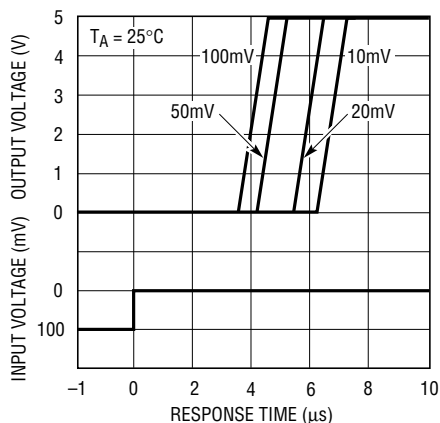
**Note 1:** Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

**Note 2:**  $IN^+ = IN^- + 80mV$ , output is in high impedance state.

**Note 3:**  $V_{CM} = 1/2(V^+ - V^-)$  for LTC1841,  $V_{CM} = V_{REF}$  for LTC1842/LTC1843.

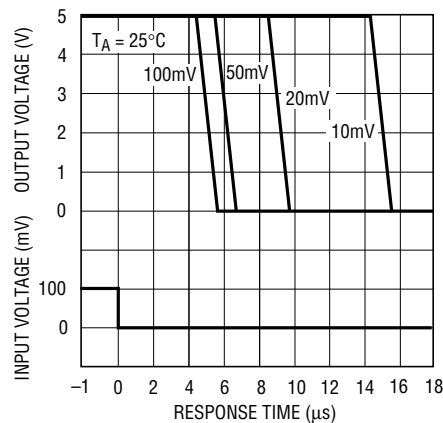
## TYPICAL PERFORMANCE CHARACTERISTICS

**Comparator Response Time vs Input Overdrive, with  $R_{PULL-UP} = 5k$**



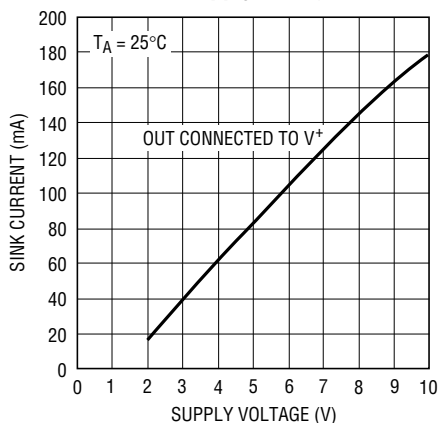
1440/1/2/3 G01

**Comparator Response Time vs Input Overdrive, with  $R_{PULL-UP} = 5k$**



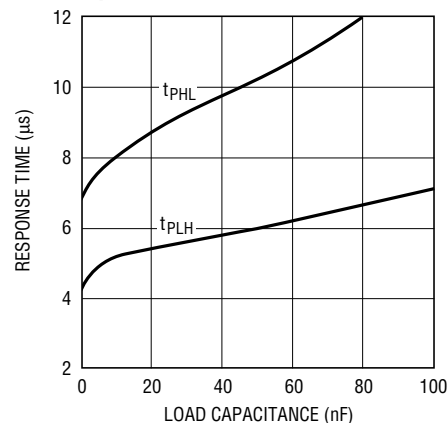
1440/1/2/3 G02

**Comparator Short-Circuit Sink Current vs Supply Voltage**



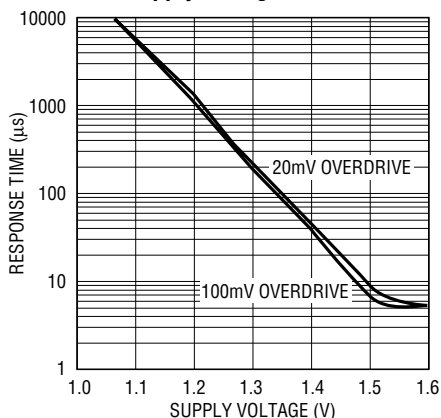
1440/1/2/3 G03

**Comparator Response Time vs Load Capacitance with 100mV Input Overdrive**



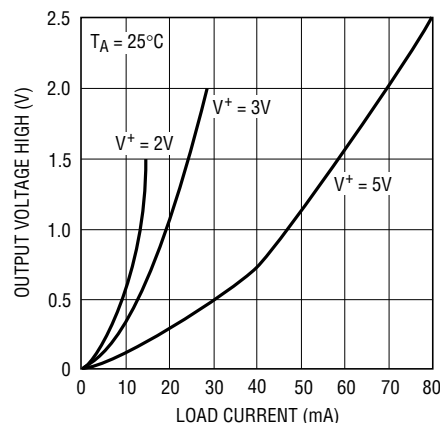
1440/1/2/3 G04

**Comparator Response Time at Low Supply Voltage**



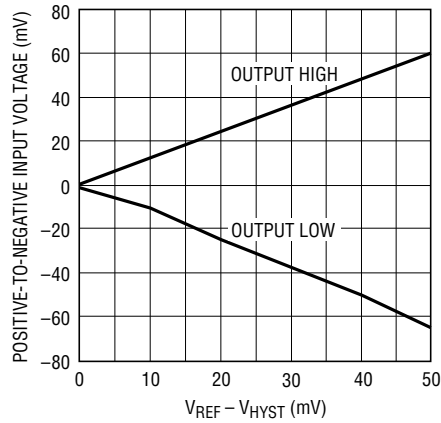
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**Comparator Output Voltage Low vs Load Current**

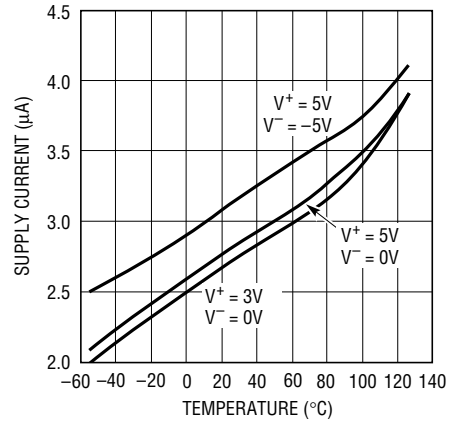


1440/1/2/3 G06

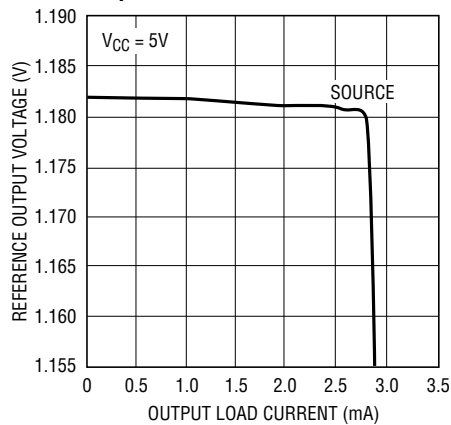
## TYPICAL PERFORMANCE CHARACTERISTICS

LTC1842/LTC1843  
Hysteresis Control

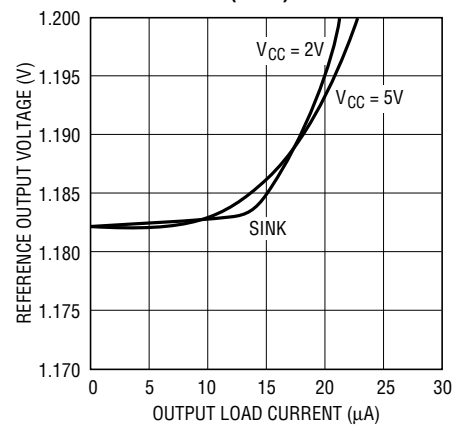
1440/1/2/3 G07

LTC1842/LTC1843  
Supply Current vs Temperature

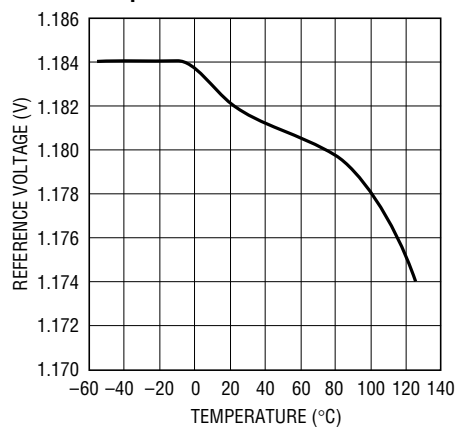
1440/1/2/3 G08

Reference Output Voltage vs  
Output Load Current

1841/2/3 G09

Reference Output Voltage vs Output  
Load Current (Sink)

1440/1/2/3 G10

Reference Voltage vs  
Temperature

1440/1/2/3 G11

## PIN FUNCTIONS

**OUT A (Pin 1):** Comparator A Open-Drain Output. Output can typically sink greater than 20mA.

**V<sup>-</sup> (Pin 2):** Negative Supply.

**IN A<sup>+</sup> (Pin 3):** Noninverting Input of Comparator A. Input common mode range extends from V<sup>-</sup> to V<sup>+</sup> – 1.3V. Input current is typically 10pA at 25°C.

**IN A<sup>-</sup> (Pin 4) (LTC1841):** Inverting Input of Comparator A. Input common mode range extends from V<sup>-</sup> to V<sup>+</sup> – 1.3V. Input current is typically 10pA at 25°C.

**IN B<sup>+</sup> (Pin 4) (LTC1842):** Noninverting Input of Comparator B. Input common mode range extends from V<sup>-</sup> to V<sup>+</sup> – 1.3V. Input current is typically 10pA at 25°C.

**IN B<sup>-</sup> (Pin 4) (LTC1843):** Inverting Input of Comparator B. Input common mode range extends from V<sup>-</sup> to V<sup>+</sup> – 1.3V. Input current is typically 10pA at 25°C.

**IN B<sup>-</sup> (Pin 5) (LTC1841):** Inverting Input of Comparator B. Input common mode range extends from V<sup>-</sup> to V<sup>+</sup> – 1.3V. Input current is typically 10pA at 25°C.

**HYST (Pin 5) (LTC1842/LTC1843):** Hysteresis Input. Connect to REF if not used. Input voltage range is from V<sub>REF</sub> to V<sub>REF</sub> – 50mV.

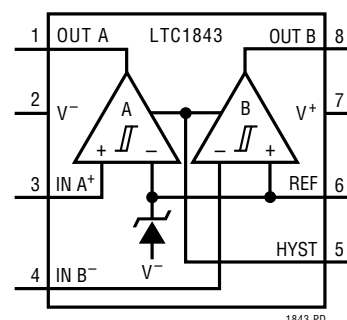
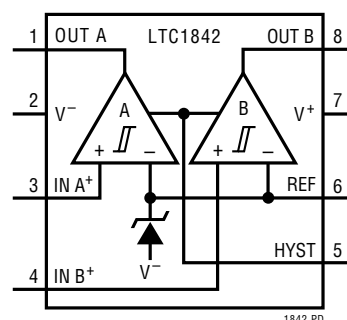
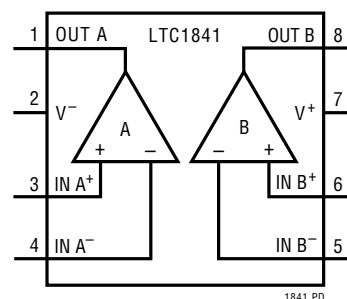
**IN B<sup>+</sup> (Pin 6) (LTC1841):** Noninverting Input of Comparator B. Input common mode range extends from V<sup>-</sup> to V<sup>+</sup> – 1.3V. Input current is typically 10pA at 25°C.

**REF (Pin 6) (LTC1842/LTC1843):** Reference Output. 1.182V with respect to V<sup>-</sup>. Can typically source greater than 1mA and sink 10μA at 25°C. Can drive 0.01μF bypass capacitor without oscillation.

**V<sup>+</sup> (Pin 7) (LTC1841):** Positive Supply. 2V to 11V.

**V<sup>+</sup> (Pin 7) (LTC1842/LTC1843):** Positive Supply. 2.5V to 11V.

**OUT B (Pin 8):** Comparator B Open-Drain Output. Output can typically sink greater than 20mA.



## APPLICATIONS INFORMATION

The LTC1841/LTC1842/LTC1843 are dual micropower comparators with a built-in 1.182V reference (LTC1842/LTC1843). Features include programmable hysteresis, wide supply voltage range (2V to 11V) and the ability for the reference to drive up to a 0.01 $\mu$ F capacitor without oscillation. The comparators' open-drain outputs can typically sink greater than 20mA and the supply current glitches that normally occur when switching logic states have been eliminated.

### Power Supplies

The comparators operate from a single 2V to 11V (2.5V to 11V for LTC1842/LTC1843) or dual  $\pm 1$ V to  $\pm 5.5$ V supply ( $\pm 1.25$ V to  $\pm 5.5$ V for LTC1842/LTC1843). If the reference output is required to source more than 1mA or the supply source impedance is high,  $V^+$  should be bypassed with a 0.1 $\mu$ F capacitor.

### Comparator Inputs

The comparators' input can swing from the negative supply  $V^-$  to within 1.3V (max) of the positive supply  $V^+$ . The input can be forced 300mV below  $V^-$  or above  $V^+$  without damage and the typical input leakage current is only  $\pm 10$ pA.

### Comparator Outputs

Each comparator output is an open-drain pull-down to  $V^-$  typically capable of sinking greater than 20mA. The low output leakage current while in three-state mode allows a high value pull-up resistor to be used. The open-drain outputs can be wire OR-ed or used in level shifting applications.

### Voltage Reference

The internal bandgap reference has an output voltage of 1.182V referenced to  $V^-$ . The reference accuracy is 1.5% from  $-40^\circ\text{C}$  to  $85^\circ\text{C}$ . It can typically source greater than 1mA and sink up to 10 $\mu$ A with a 5V supply. The reference can drive a bypass capacitor of up to 0.01 $\mu$ F without oscillation. By inserting a series resistor, capacitance values up to 100 $\mu$ F can be used (Figure 1).

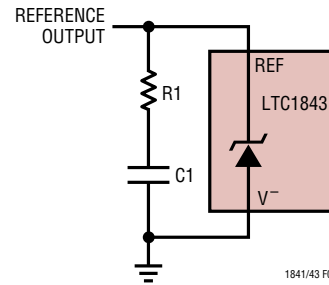


Figure 1. Damping the Reference Output

Figure 2 shows the resistor value required for different capacitor values to achieve critical damping. Bypassing the reference can help prevent false tripping of the comparators by preventing glitches on  $V^+$  or reference load transients from disturbing the reference output voltage.

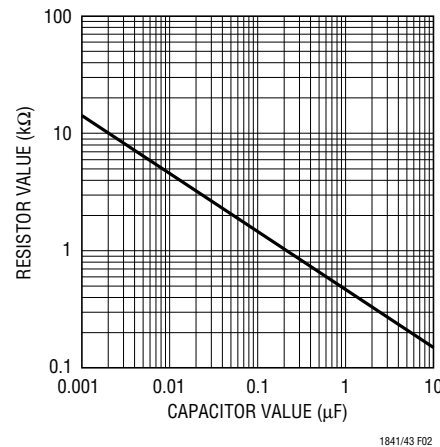
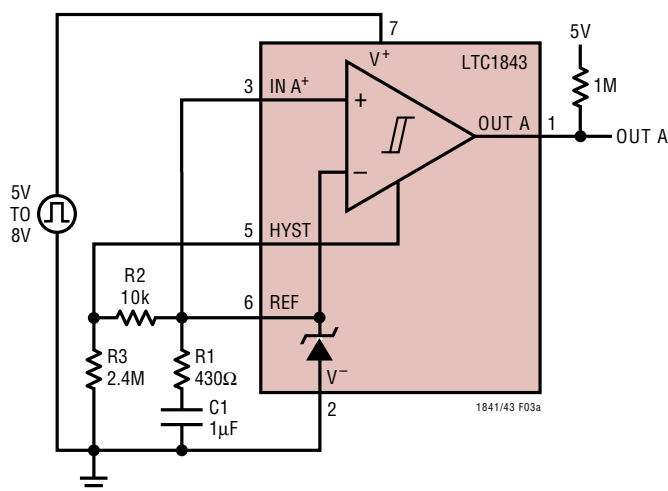


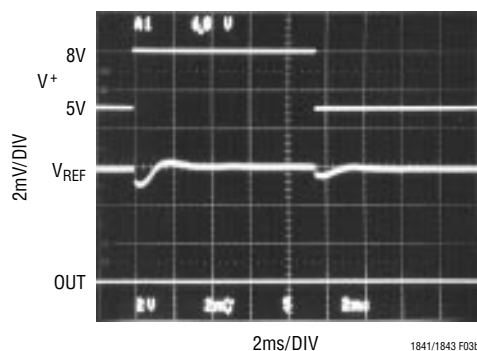
Figure 2. Damping Resistance vs Bypass Capacitor Value

Figure 3 shows the bypassed reference output with a square wave applied to the  $V^+$  pin. Resistors R2 and R3 set a 10mV hysteresis voltage band while R1 damps the reference response. Note that the comparator output doesn't trip.

## APPLICATIONS INFORMATION



### Figure 3a. Power Supply Transient Test Circuit



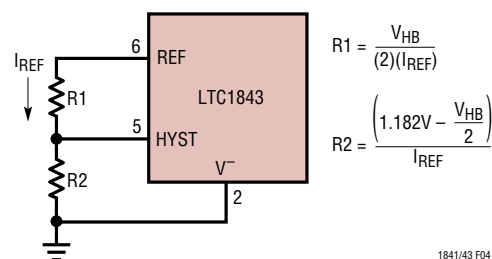
### Figure 3b. Power Supply Transient Rejection

## Hysteresis

Hysteresis can be added to the LTC1842/LTC1843 by connecting a resistor (R1) between the REF and HYST pins and a second resistor (R2) from HYST to V<sup>-</sup> (Figure 4).

The difference between the upper and lower threshold voltages, or hysteresis voltage band ( $V_{HB}$ ), is equal to twice the voltage difference between the REF and HYST pins.

As more hysteresis is added, the upper threshold increases the same amount as the low threshold decreases. The maximum voltage allowed between REF and HYST pins is 50mV, producing a maximum hysteresis voltage band of 100mV. The hysteresis band may vary by up to

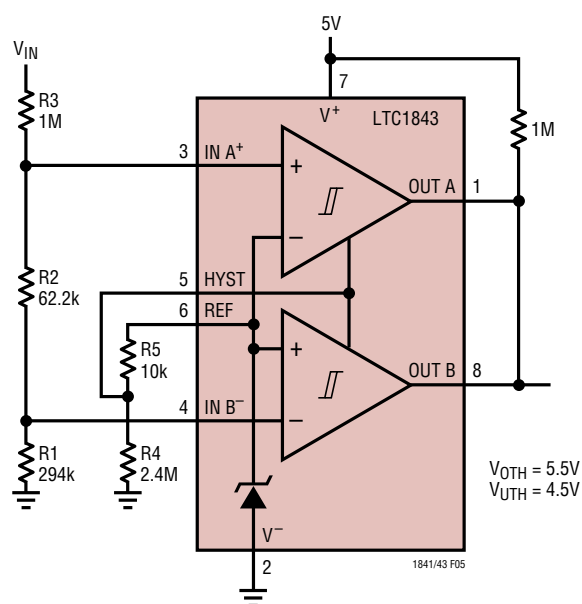


### Figure 4. Programmable Hysteresis

15%. If hysteresis is not wanted, the HYST pin should be shorted to REF. Acceptable values for  $I_{REF}$  range are from  $0.1\mu A$  to  $5\mu A$ . If  $2.4M$  is chosen for  $R_2$ , then the value of  $R_1$  is equal to the value of  $V_{HR}$ .

## Window Detector

The LTC1843 is ideal for use as a micropower window detector as shown in Figure 5. The values of R1, R2 and R3 are selected for a 4.5V undervoltage threshold and a 5.5V overvoltage threshold. R4 and R5 set the hysteresis voltage. The following design procedure can be used to select the component values:



### Figure 5. Dual Level Detector



## APPLICATIONS INFORMATION

1. Choose the required hysteresis voltage band and calculate values for R4 and R5 according to the formulas in the hysteresis section. In this example,  $\pm 5\text{mV}$  of hysteresis has been added at the comparator input ( $V_H = V_{HB}/2$ ). Note that the hysteresis apparent at  $V_{IN}$  will be larger because of the input resistor divider.
2. Select R1. The leakage current into  $\text{IN } B^-$  is under  $1\text{nA}$  so the current through R1 should exceed  $100\text{nA}$ , to ensure threshold accuracy. R1 values up to about  $10\text{M}$  can be used, but values in the  $100\text{k}$  to  $1\text{M}$  range are usually easier to deal with. In this example choose  $R1 = 294\text{k}$ .

3. Calculate  $R2 + R3$ . The overvoltage threshold should be set at  $5.5\text{V}$ . The design equation is as follows:

$$\begin{aligned} R2 + R3 &= R1 \left( \frac{V_{UTH}}{V_{REF} + V_H} - 1 \right) \\ &= 294\text{k} \left( \frac{5.5}{1.182 + 0.005} - 1 \right) \\ &= 1.068\text{M} \end{aligned}$$

4. Calculate R2. The undervoltage threshold should be set at  $4.5\text{V}$ . The design equation is as follows:

$$\begin{aligned} R2 &= (R1 + R2 + R3) \frac{V_{REF} - V_H}{V_{LTH}} - R1 \\ &= (294\text{k} + 1.068\text{M}) \frac{1.182 - 0.005}{4.5} - 294\text{k} \\ &= 62.2\text{k} \end{aligned}$$

Choose  $R2 = 61.9\text{k}$  (1% standard value)

5. Calculate R3:

$$\begin{aligned} R3 &= (R2 + R3) - R2 \\ &= 1.068\text{M} - 61.9\text{k} \\ &= 1.006\text{M} \end{aligned}$$

Choose  $R3 = 1\text{M}$  (1% standard value)

6. Verify the resistor values. The equations are as follows, evaluated for the above example:

Overvoltage threshold:

$$\begin{aligned} V_{OTH} &= (V_{REF} + V_H) \frac{R1 + R2 + R3}{R1} \\ &= 5.474\text{V} \end{aligned}$$

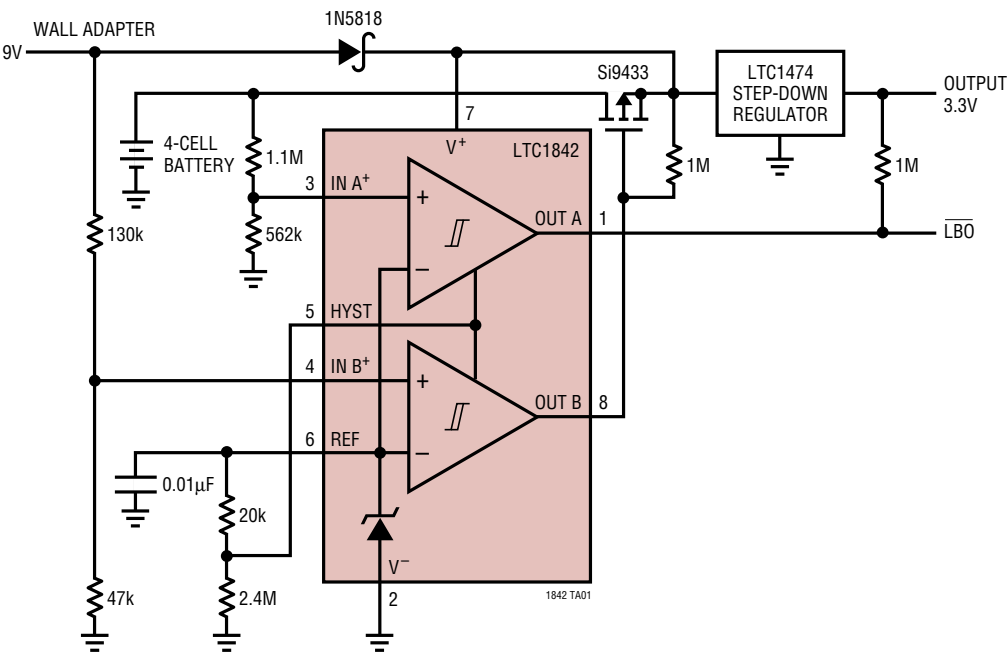
Undervoltage threshold:

$$\begin{aligned} V_{UTH} &= (V_{REF} - V_H) \frac{R1 + R2 + R3}{R1 + R2} \\ &= 4.484\text{V} \end{aligned}$$

where the hysteresis voltage  $V_H = (V_{REF}) \left( \frac{R5}{R4} \right)$

TYPICAL APPLICATION

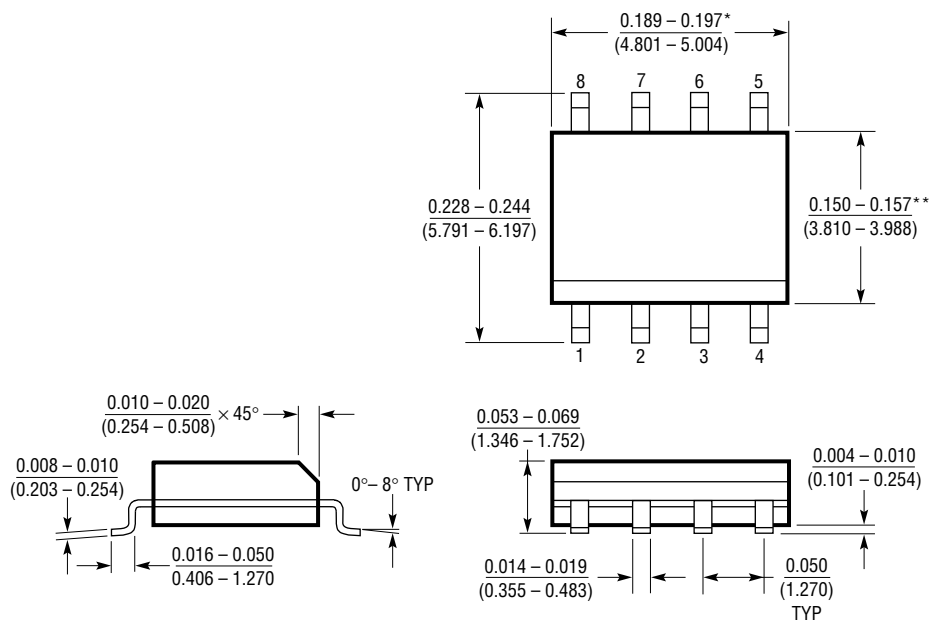
Battery Switchover Circuit



**PACKAGE DESCRIPTION**

Dimensions in inches (millimeters) unless otherwise noted.

**S8 Package**  
**8-Lead Plastic Small Outline (Narrow 0.150)**  
 (LTC DWG # 05-08-1610)



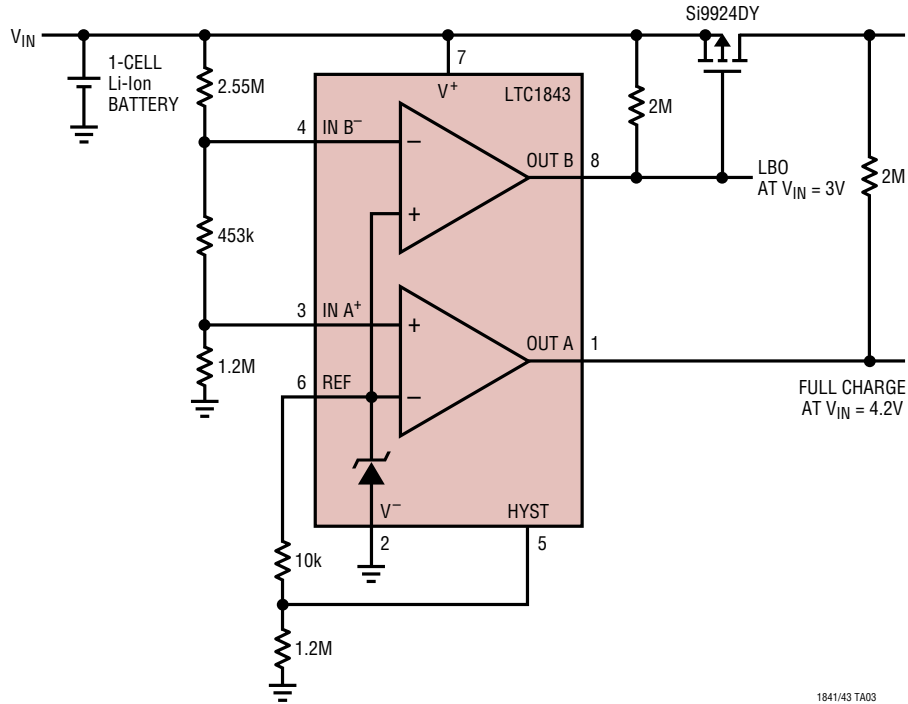
\*DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

\*\*DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE

S08 0996

## TYPICAL APPLICATION

Low-Battery Load Disconnect and Charge Termination



1841/43 TA03

## RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT <sup>®</sup> 1178/LT1179	Dual/Quad 17 $\mu$ A Precision Single Supply Op Amps	70 $\mu$ V Max $V_{OS}$ , 5nA Max $I_{BIAS}$
LT1351	Single 250 $\mu$ A, 3MHz, 200V/ $\mu$ s Op Amp with Shutdown	C-Load <sup>™</sup> Op Amp Stable Driving Any Capacitive Load
LT1352/LT1353	Dual/Quad 250 $\mu$ A, 3MHz, 200V/ $\mu$ s Op Amps	C-Load Op Amps Stable Driving Any Capacitive Load
LTC1440/LTC1540	Micropower Comparator with 1% Reference	1.182V $\pm$ 1% Reference, $\pm$ 10mV (Max) Input Offset
LTC1441/LTC1442	Micropower Dual Comparator with 1% Reference	1.182V $\pm$ 1% Reference (LTC1442)
LTC1443/LTC1444/LTC1445	Micropower Quad Comparator with 1% Reference	LTC1443 Has 1.182V Reference, LTC1444/LTC1445 Have 1.221V Reference and Adjustable Hysteresis
LTC1474	Low Quiescent Current High Efficiency Step-Down Switching Regulator	10 $\mu$ A Standby Current, 92% Efficiency, Space Saving 8-Pin MSOP Package
LT1495	1.5 $\mu$ A Max, Dual Precision Rail-to-Rail Input and Output Op Amp	375 $\mu$ V Max $V_{OS}$ , 250pA $I_{BIAS}$ , 25pA $I_{OS}$
LT1521	300mA Low Dropout Regulator with Micropower Quiescent Current and Shutdown	0.5V Dropout Voltage, 12 $\mu$ A Quiescent Current, Adjustable Output 3V, 3.3V and 5V Fixed
LTC1541/LTC1542	Micropower Op Amp, Comparator and Reference	1.200V $\pm$ 0.8% Reference (LTC1541) Op Amp Outputs Stable with 1000pF Load
LT1634	Micropower Precision Shunt Voltage Reference	1.25V Output, 10 $\mu$ A Operating Current, 0.1% Initial Accuracy 10ppm/ $^{\circ}$ C Max Drift

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